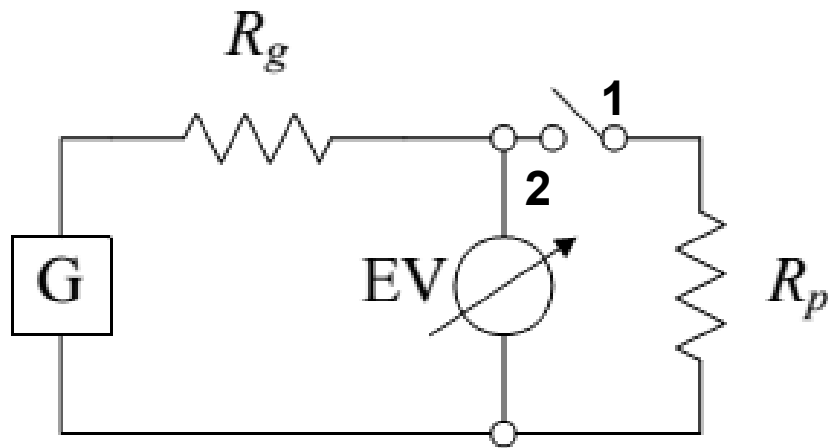


Merenje električnih veličina II

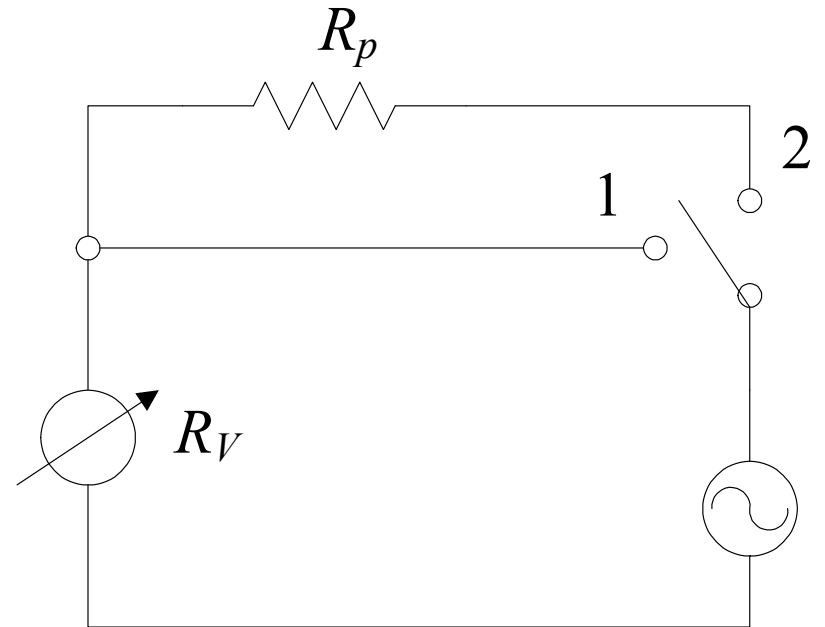
12.11.2020.

MERENJE IZLAZNE OTPORNOSTI
GENERATORA (lab. vežba)



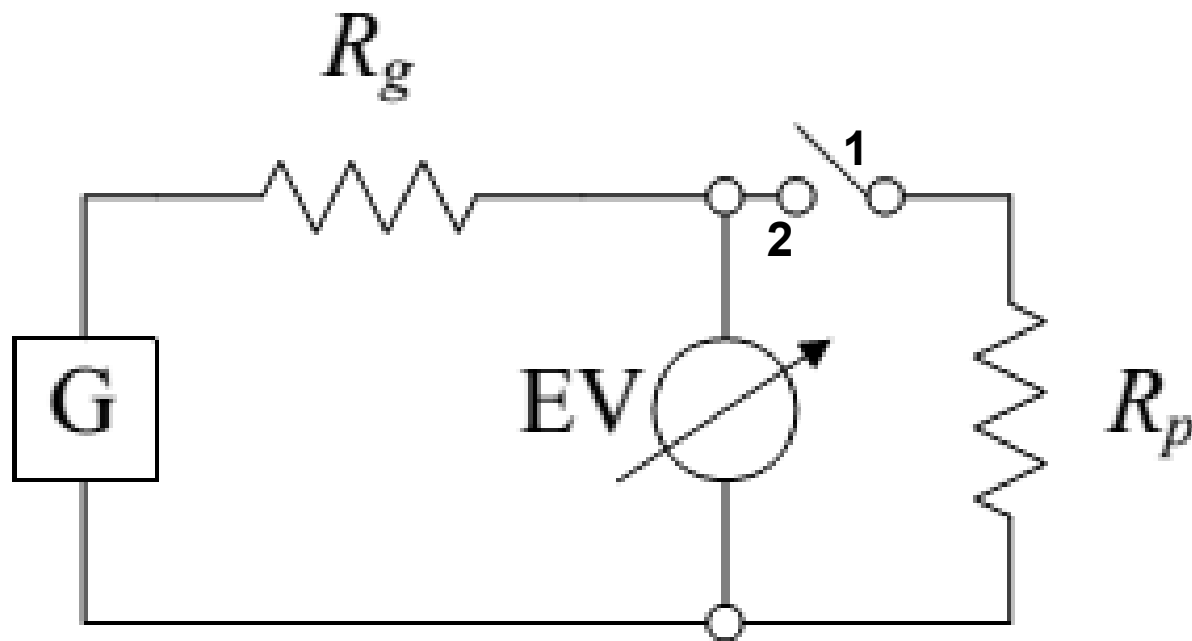
$$R_g = f(R_p, U_1, U_2) = ?$$

MERENJE ULAZNE
OTPORNOSTI VOLTMETRA
(lab. vežba)



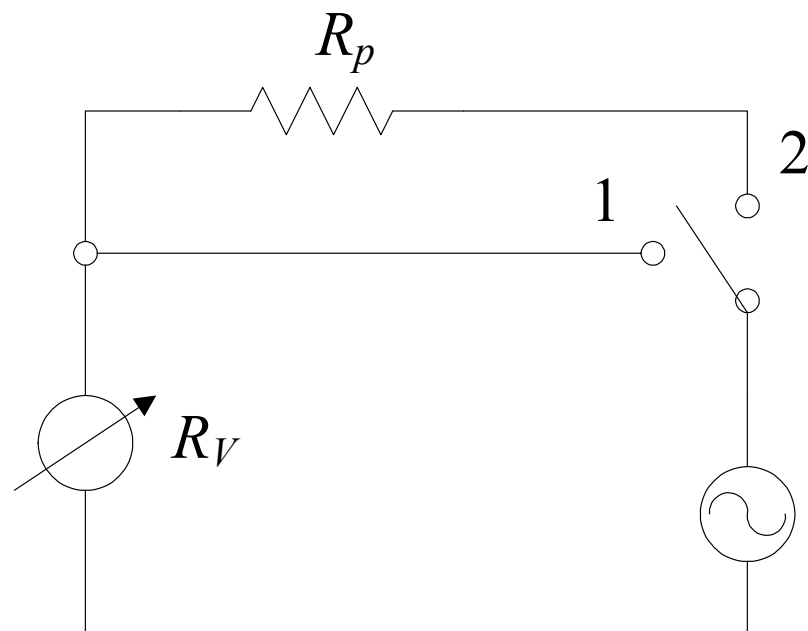
$$R_V = f(R_p, U_1, U_2) = ?$$

MERENJE IZLAZNE OTPORNOSTI GENERATORA (lab. vežba)



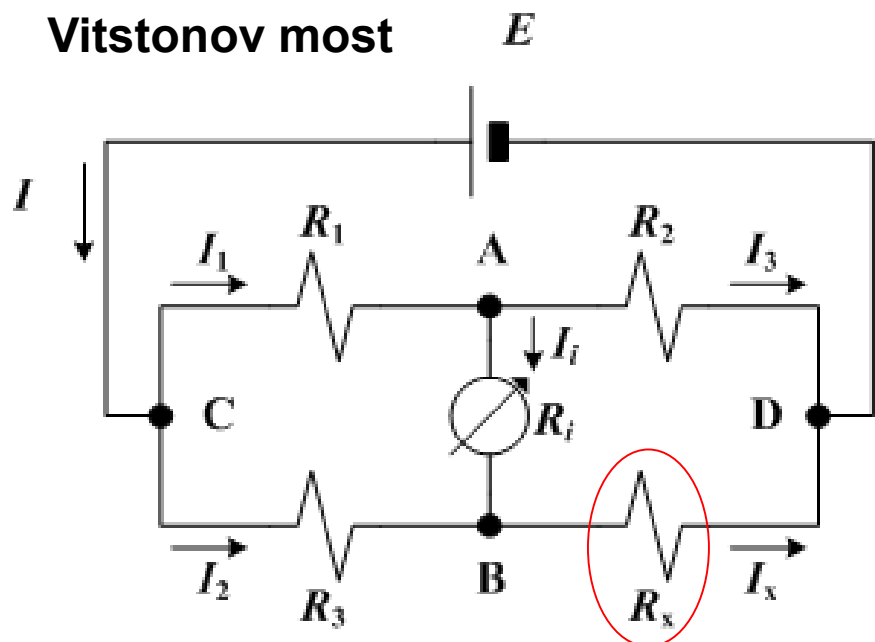
$$R_g = R_p \frac{U_1 - U_2}{U_2}$$

MERENJE ULAZNE OTPORNOSTI VOLTMETRA (lab. vežba)



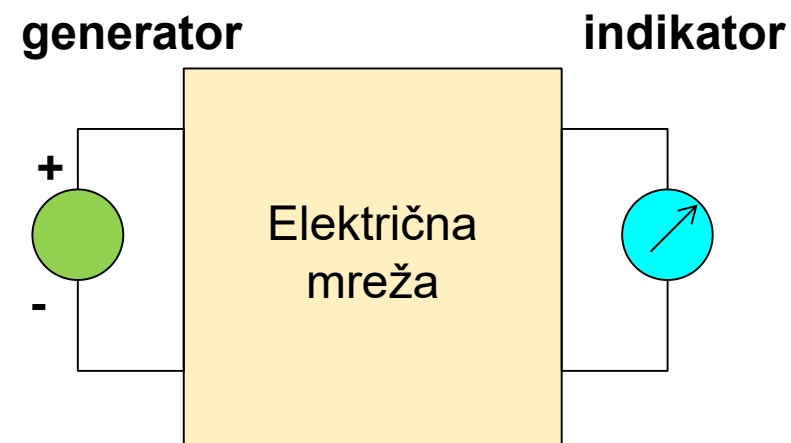
$$R_V = R_p \frac{U_2}{U_1 - U_2}$$

MERENJE OTPORNOSTI MOSTOM



Primer mosta za merenje otpornosti R_x

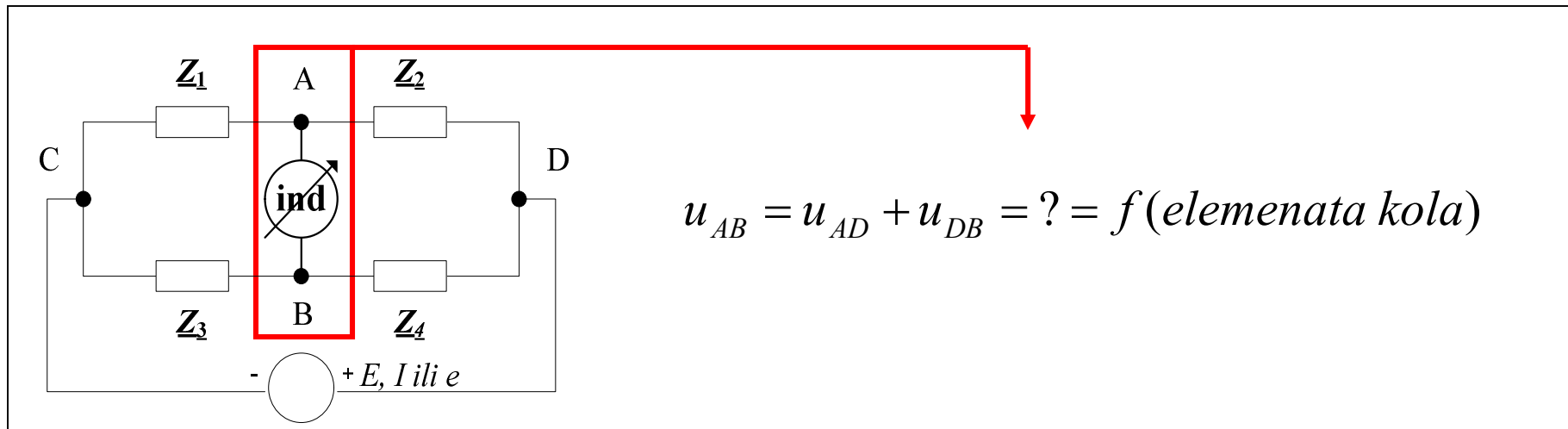
1847. - "Otporna vaga"



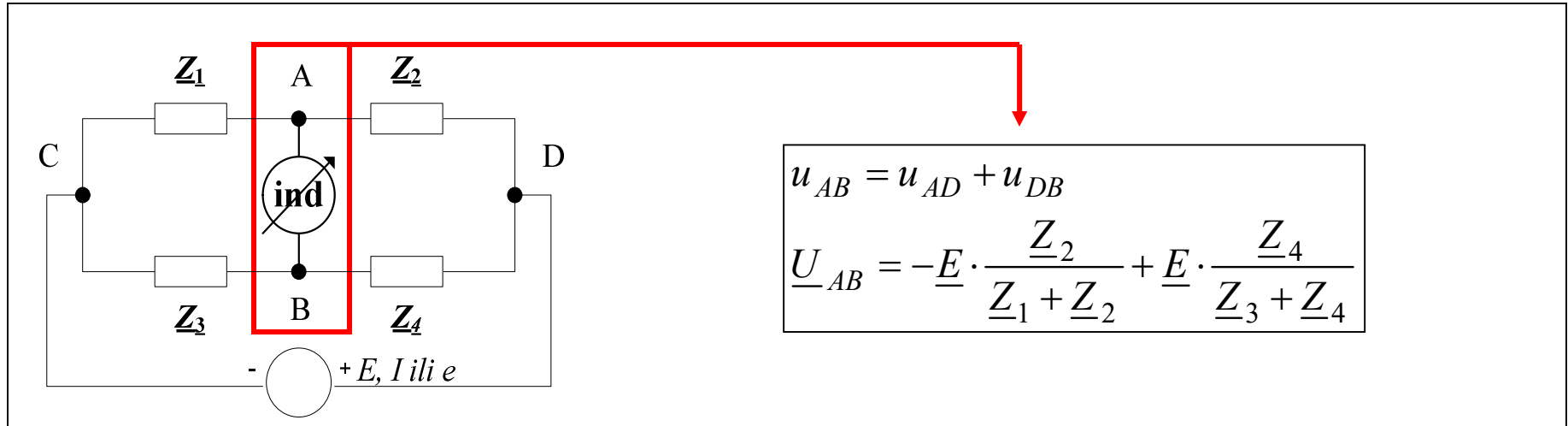
Opšta konfiguracija mernog mosta

Indikator (nul-detektor) može biti:
galvanometar,
osetljiv ampermetar,
voltmetar,
osciloskop,
slušalice

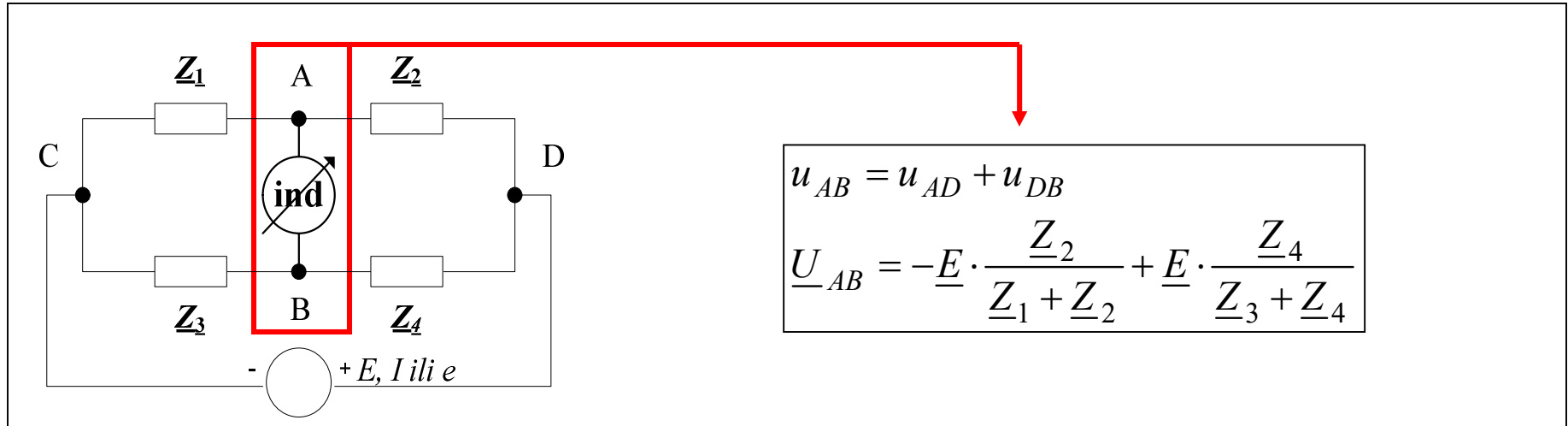
OPŠTA KONFIGURACIJA MERNOG MOSTA



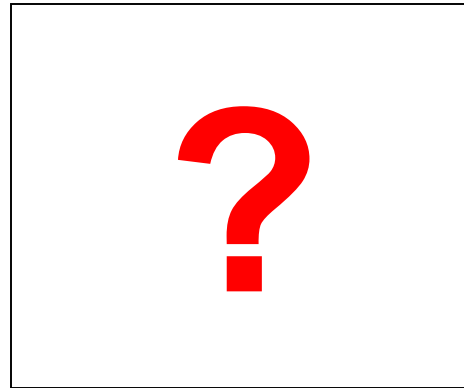
OPŠTA KONFIGURACIJA MERNOG MOSTA



OPŠTA KONFIGURACIJA MERNOG MOSTA



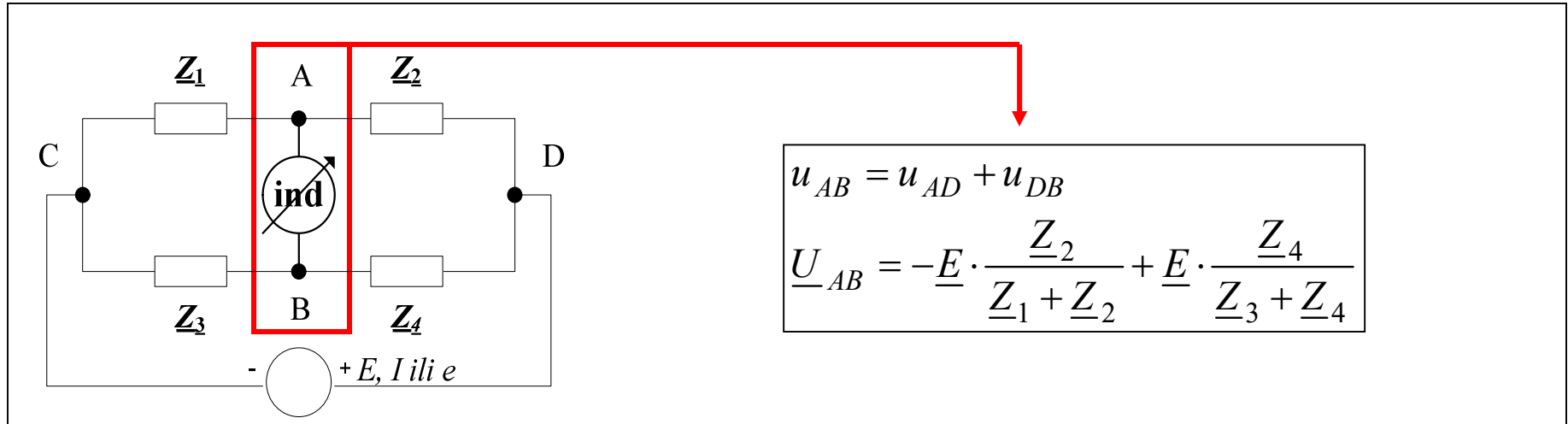
u_{AB}



0

= ili \neq ?

OPŠTA KONFIGURACIJA MERNOG MOSTA



uravnotežen:

cilj: $u_{AB} = 0$

$\Rightarrow \underline{Z}_1 \underline{Z}_4 = \underline{Z}_2 \underline{Z}_3$

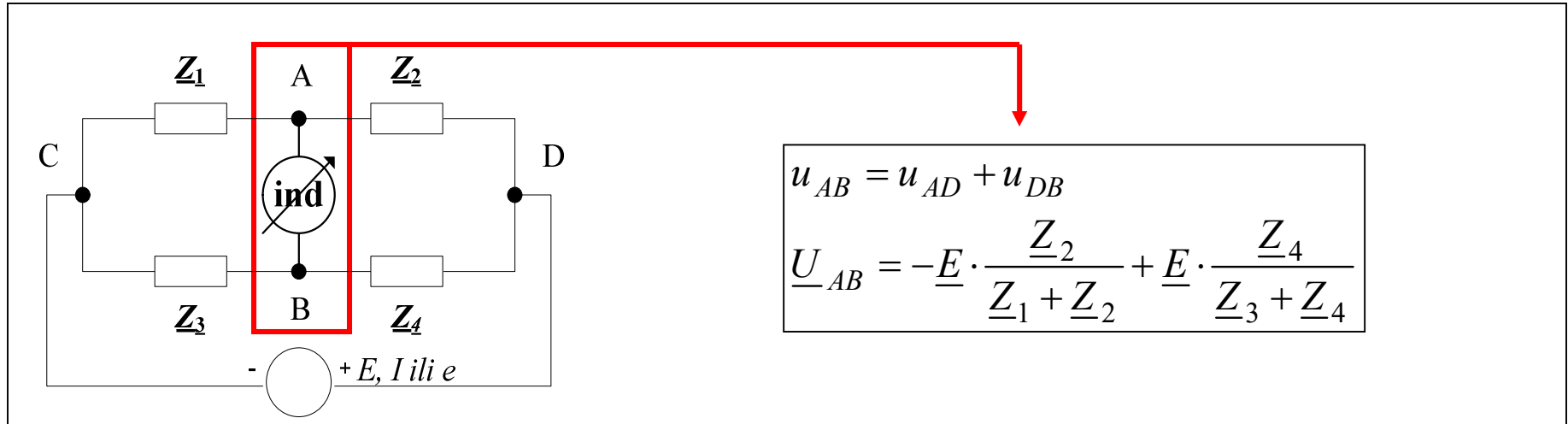
ova relacija važi
SAMO za
uravnotežen most!!!

MOST

neuravnotežen:

cilj: $u_{AB} \neq 0$

OPŠTA KONFIGURACIJA MERNOG MOSTA



uravnotežen most
PRIMENA

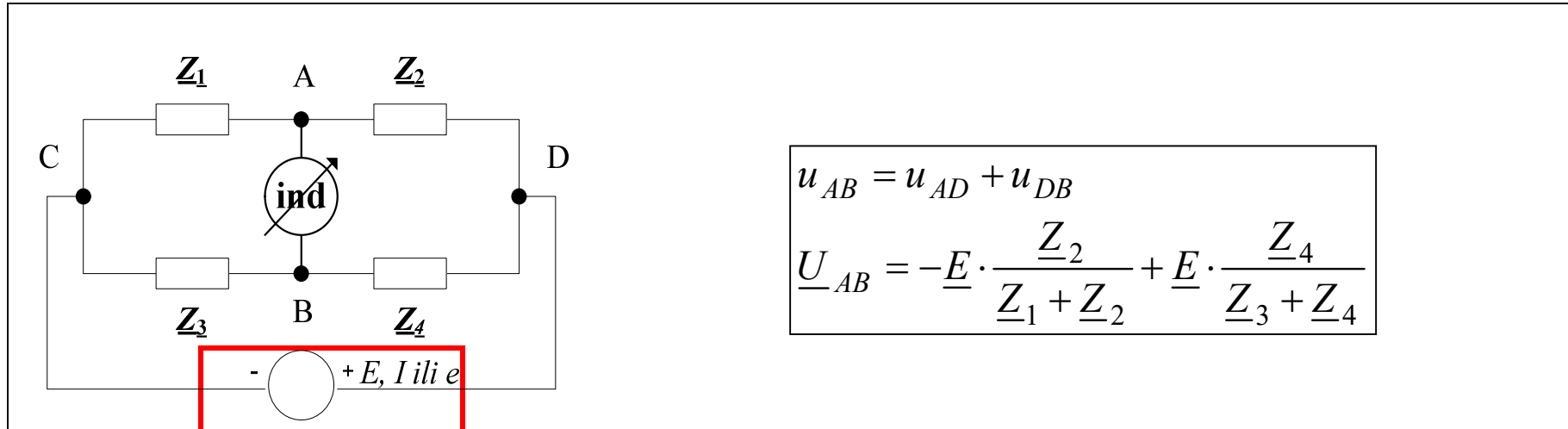
Npr.
Određivanje
nepoznatog
elementa kola

MOST

neuravnotežen most
PRIMENA

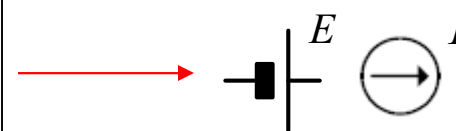
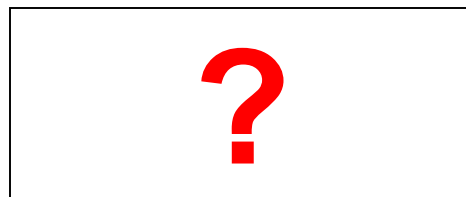
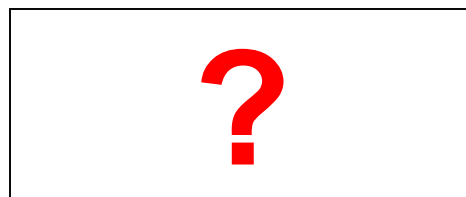
Npr.
Određivanje
promene
nepoznatog
elementa kola

OPŠTA KONFIGURACIJA MERNOG MOSTA



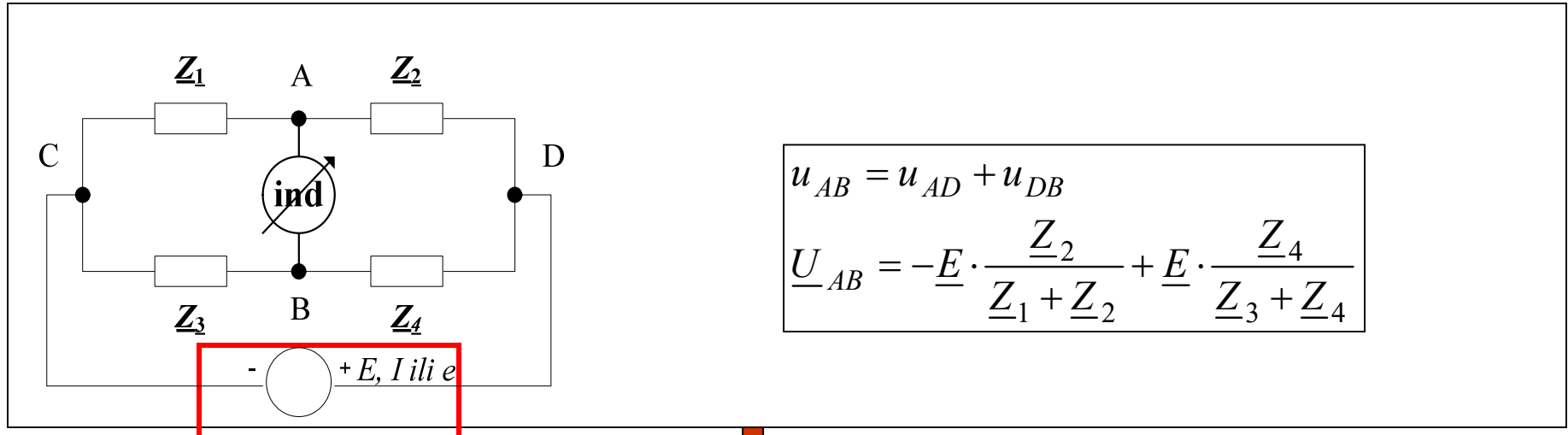
$$u_{AB} = u_{AD} + u_{DB}$$

$$\underline{U}_{AB} = -\underline{E} \cdot \frac{\underline{Z}_2}{\underline{Z}_1 + \underline{Z}_2} + \underline{E} \cdot \frac{\underline{Z}_4}{\underline{Z}_3 + \underline{Z}_4}$$



E ili \neq e ?

OPŠTA KONFIGURACIJA MERNOG MOSTA



$$u_{AB} = u_{AD} + u_{DB}$$

$$\underline{U}_{AB} = -\underline{E} \cdot \frac{\underline{Z}_2}{\underline{Z}_1 + \underline{Z}_2} + \underline{E} \cdot \frac{\underline{Z}_4}{\underline{Z}_3 + \underline{Z}_4}$$

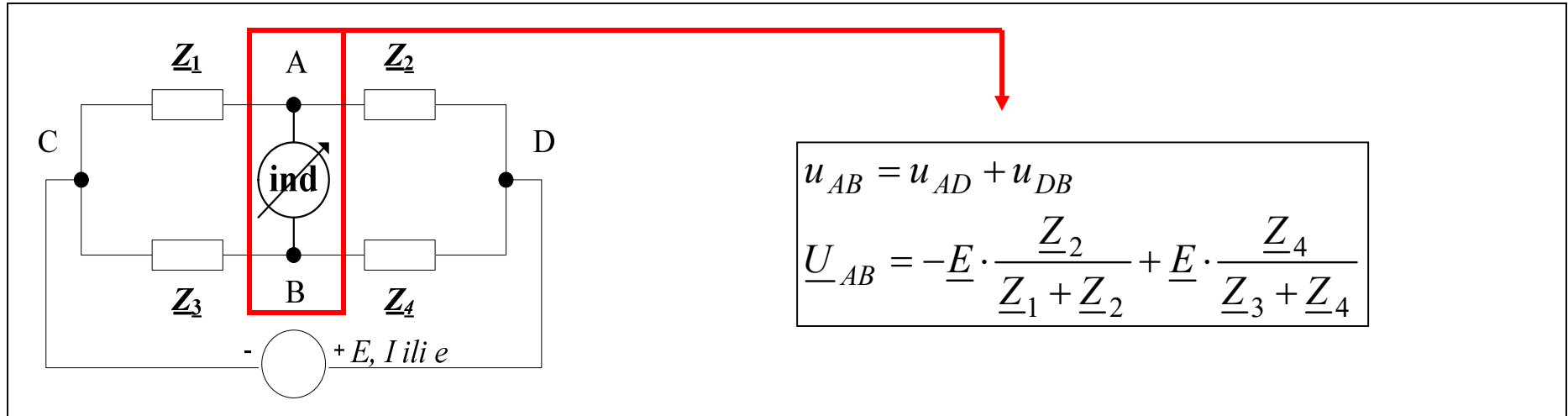


MOST

za jednosmernu struju:
 • Z_i su otpornosti
 • DC generator

za naizmeničnu struju:
 • Z_i su paralelne i/ili redne veze otpornosti, induktivnosti, kapacitivnosti
 • AC generator

OPŠTA KONFIGURACIJA MERNOG MOSTA

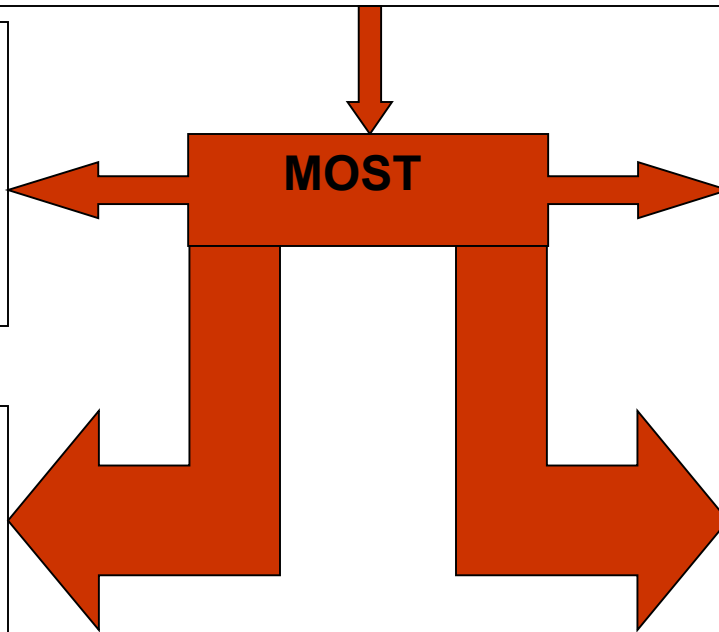


uravnotežen:

cilj: $u_{AB} = 0$
 $\Rightarrow \underline{Z}_1 \underline{Z}_4 = \underline{Z}_2 \underline{Z}_3$

neuravnotežen:

cilj: $u_{AB} \neq 0$



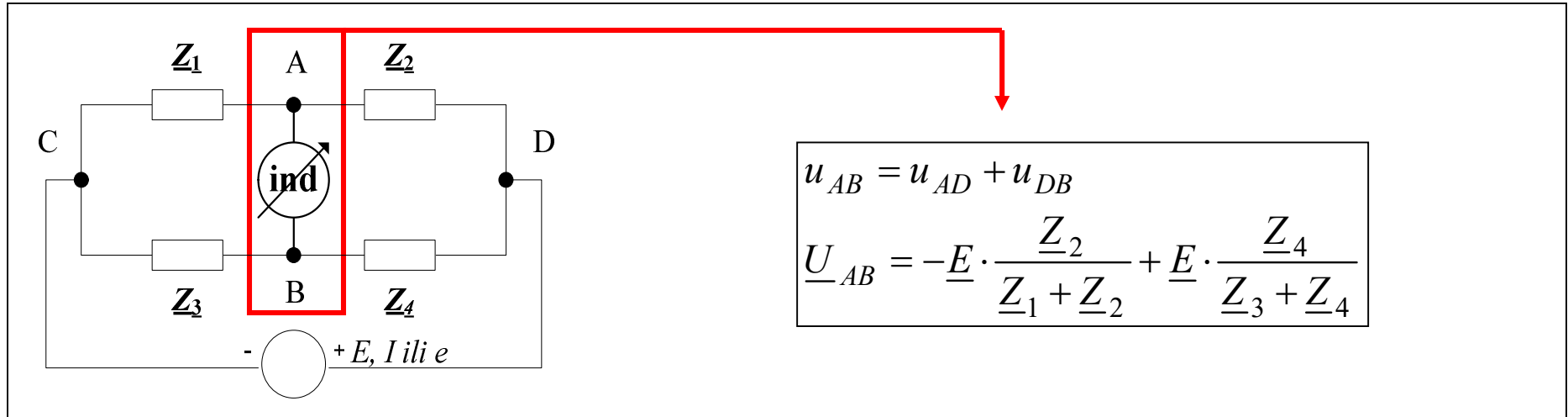
za jednosmernu struju:

- Z_i su otpornosti
- DC generator

za naizmničnu struju:

- Z_i su paralelne i/ili redne veze otpornosti, induktivnosti, kapacitivnosti
- AC generator

OPŠTA KONFIGURACIJA MERNOG MOSTA



Osetljivost mosta je osnovna karakteristika mosta.

Od interesa je vrednost osetljivosti u blizini ravnoteže mosta.

$$S_U = \frac{\partial \underline{U}_{AB}}{\partial \underline{Z}_x} = \underline{Z}_x \frac{\partial \underline{U}_{AB}}{\partial \underline{Z}_x}$$

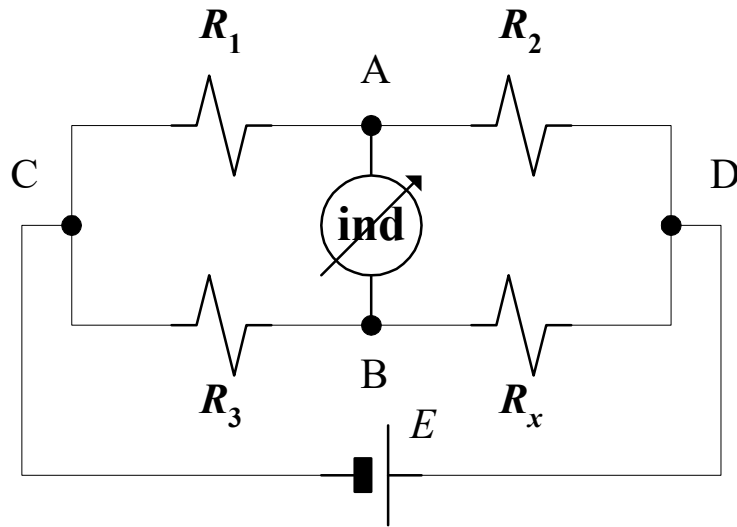
Naponska osetljivost mosta

$$S_I = \frac{\partial \underline{I}_{AB}}{\partial \underline{Z}_x}$$

Strujna osetljivost mosta

MERENJE OTPORNOSTI MOSTOM

Vitstonov most - **uravnoteženi**



Odrediti nepoznatu otpornost R_x u funkciji elemenata kola. Smatrati da je indikator idealan voltmetar.

Za uravnotežen most važi $\Rightarrow R_1 R_x = R_2 R_3$

$$R_x = \frac{R_2 R_3}{R_1}$$

Za određivanje nepoznate otpornosti R_x je potrebno poznavati jedan od otpornika i odnos preostala dva otpornika

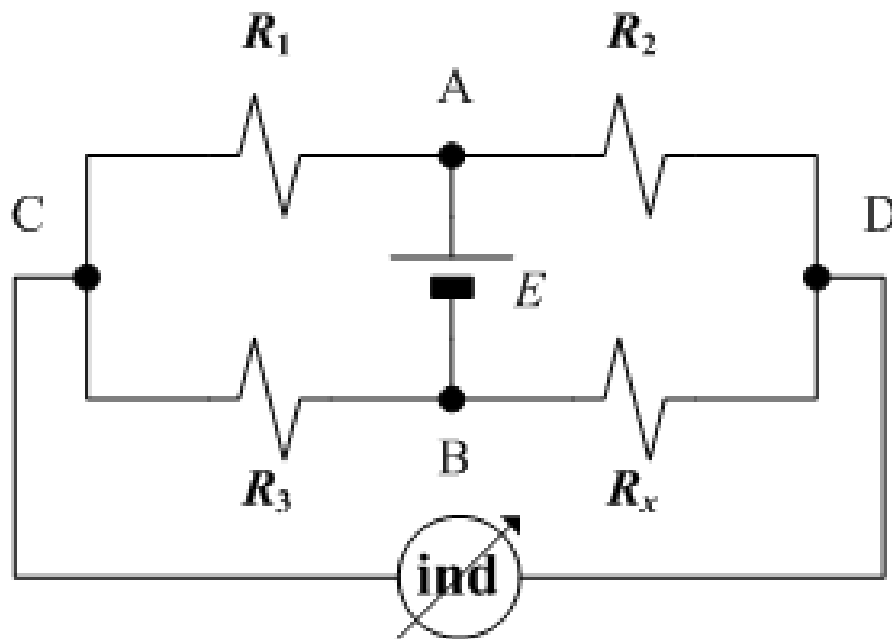
Napajanje (napon i impedansa) ne utiče na vrednost R_x – **voditi računa o maksimalnoj snazi otpornika!**

Osetljivost i izlazna impedansa nul-detektora ne utiču na vrednost R_x

Merni opseg: od 1Ω (zbog otpornosti pristupnih vodova) do $10 \text{ M}\Omega$.

MERENJE OTPORNOSTI MOSTOM

Vitstonov most - uravnoveženi

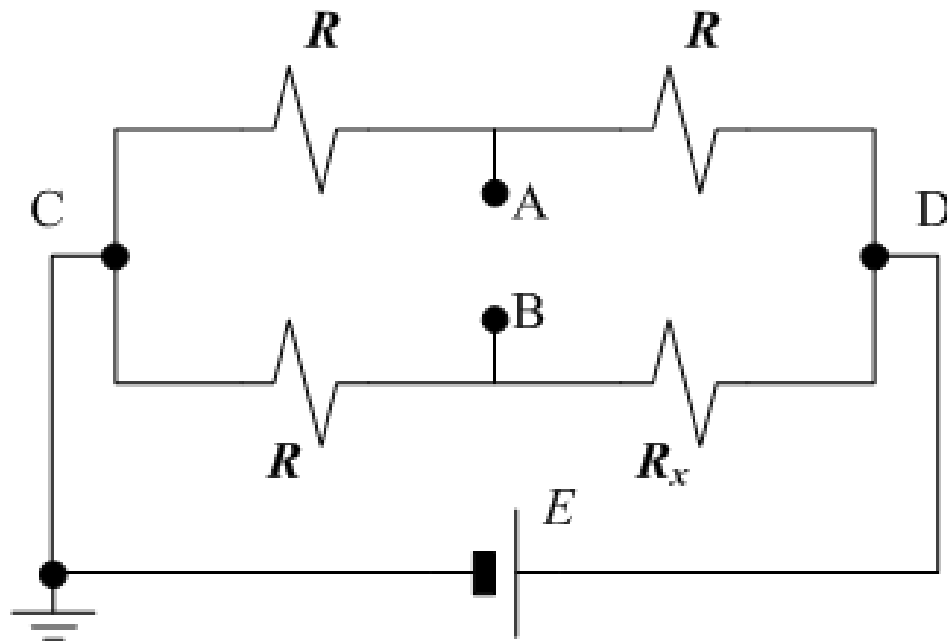


Koji je uslov ravnoteže?

MERENJE OTPORNOSTI MOSTOM

Vitstonov most - uravnoteženi

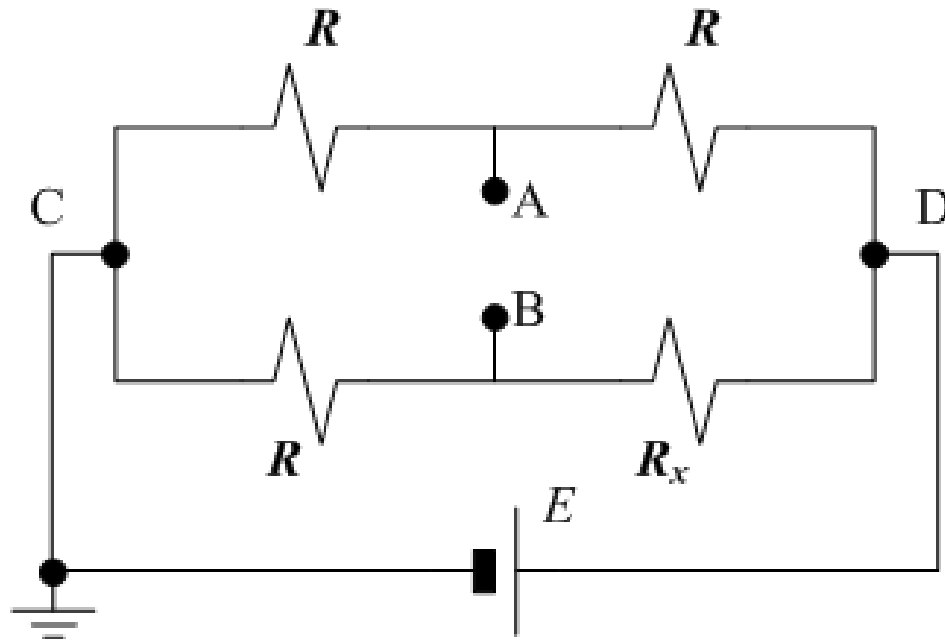
Kako povezati osciloskop kao nul-detektor?



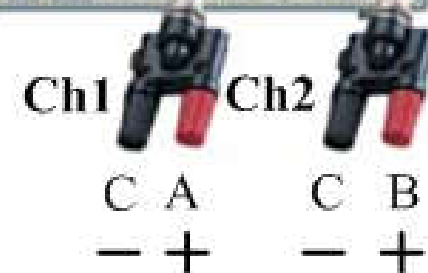
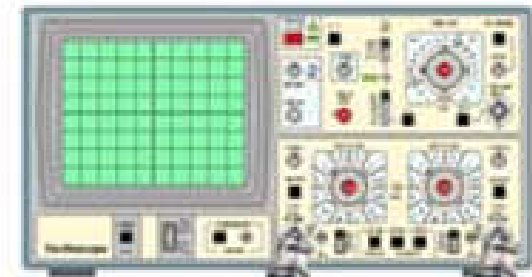
MERENJE OTPORNOSTI MOSTOM

Vitstonov most - uravnoteženi

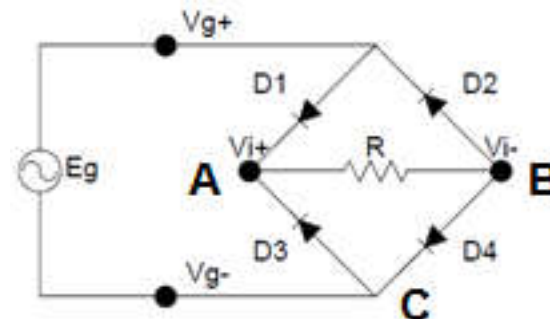
Kako povezati osciloskop kao nul-detektor?



Ch1 (norm)
Ch2 (inv)
add

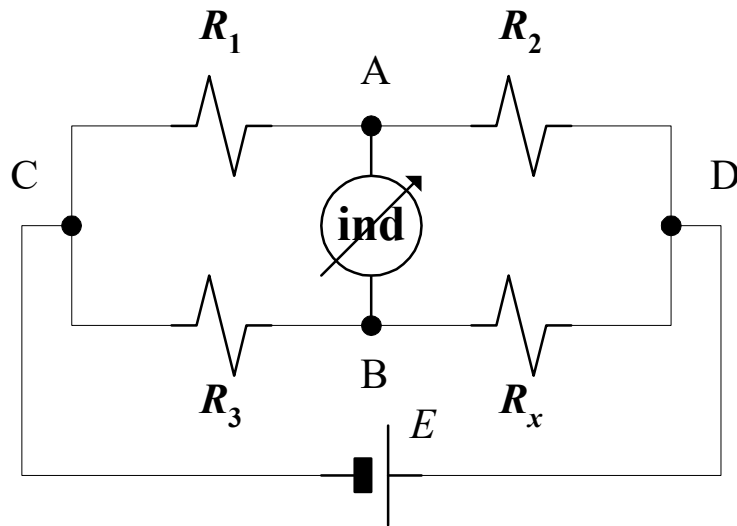


Analogija: kolo za dvostrano ispravljanje



MERENJE OTPORNOSTI MOSTOM

Vitstonov most - uravnoveženi



Kolika je naponska osetljivost mosta?

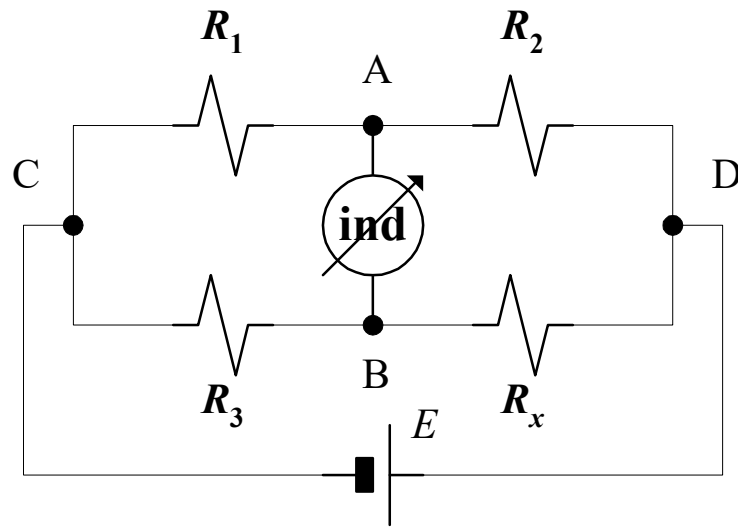
$$u_{AB} = u_{AD} + u_{DB} = -E \cdot \frac{R_2}{R_1 + R_2} + E \cdot \frac{R_x}{R_x + R_3}$$

$$S_U = R_x \frac{\partial u_{AB}}{\partial R_x} = E \frac{R_3 R_x}{(R_x + R_3)^2}$$

$$R_x = \frac{R_2 R_3}{R_1} \quad \Rightarrow \quad S_U = E \frac{\frac{R_2}{R_1}}{\left(1 + \frac{R_2}{R_1}\right)^2} \quad \Rightarrow \quad S_{U \max} = \frac{E}{4}, \quad \text{za } R_1 = R_2$$

MERENJE OTPORNOSTI MOSTOM

Vitstonov most – neuravnoteženi



Odrediti promenu nepoznate otpornosti R_x u funkciji elemenata kola i napona u_{AB} . Smatrati da je indikator idealan voltmetar.

Poznato je da:

$$R_1 = R_2 = R_3 = R$$
$$R_x = R \pm \Delta R_x, \Delta R_x / R \ll 1$$

(u praksi $\Delta R_x / R = 5\%$)

$$u_{AB} = u_{AD} + u_{DB} = -E \cdot \frac{R_2}{R_1 + R_2} + E \cdot \frac{R_x}{R_x + R_3} = E \cdot \frac{R_x R_1 - R_2 R_3}{(R_1 + R_2)(R_x + R_3)}$$

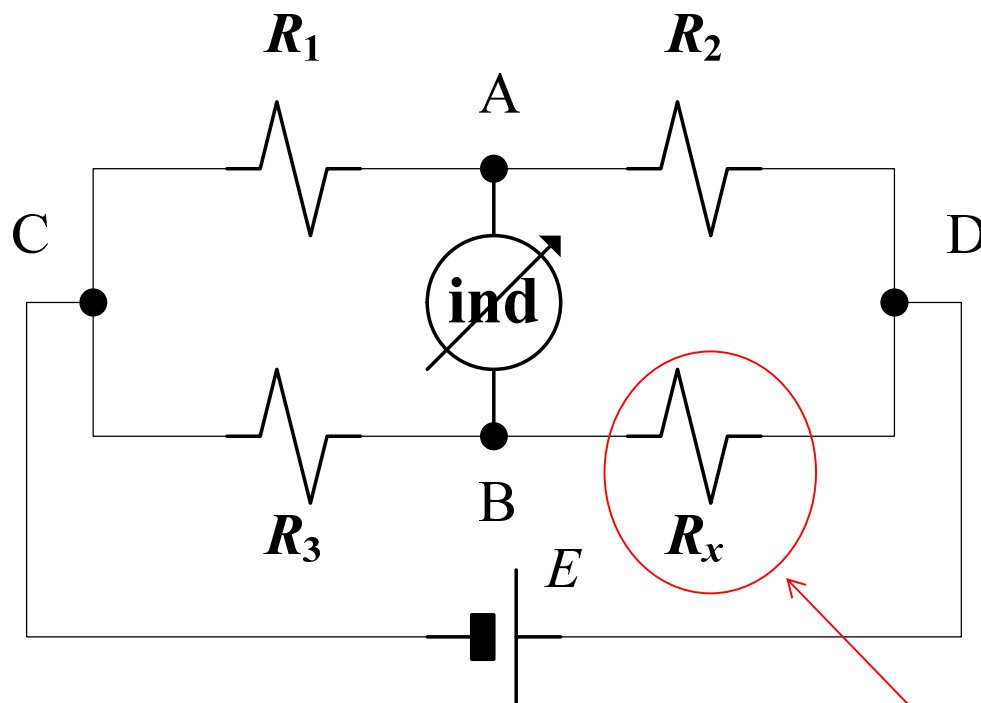
$$u_{AB} = E \cdot \frac{(R + \Delta R_x)R - R^2}{2R(R + \Delta R_x + R)} = E \cdot \frac{\Delta R_x}{2R \left(2 + \frac{\Delta R_x}{R} \right)} \approx E \cdot \frac{\Delta R_x}{4R}$$

$$\Rightarrow \Delta R_x = \frac{4R}{E} \cdot u_{AB}$$

Merni opseg za merenja neuravnoteženim Vitstonovim mostom je: $R \pm \Delta R_x$

MERENJE OTPORNOSTI MOSTOM

Vitstonov most – **neuravnoteženi** - PRIMENA



$$\Delta R_x = \frac{4R}{E} \cdot u_{AB}$$

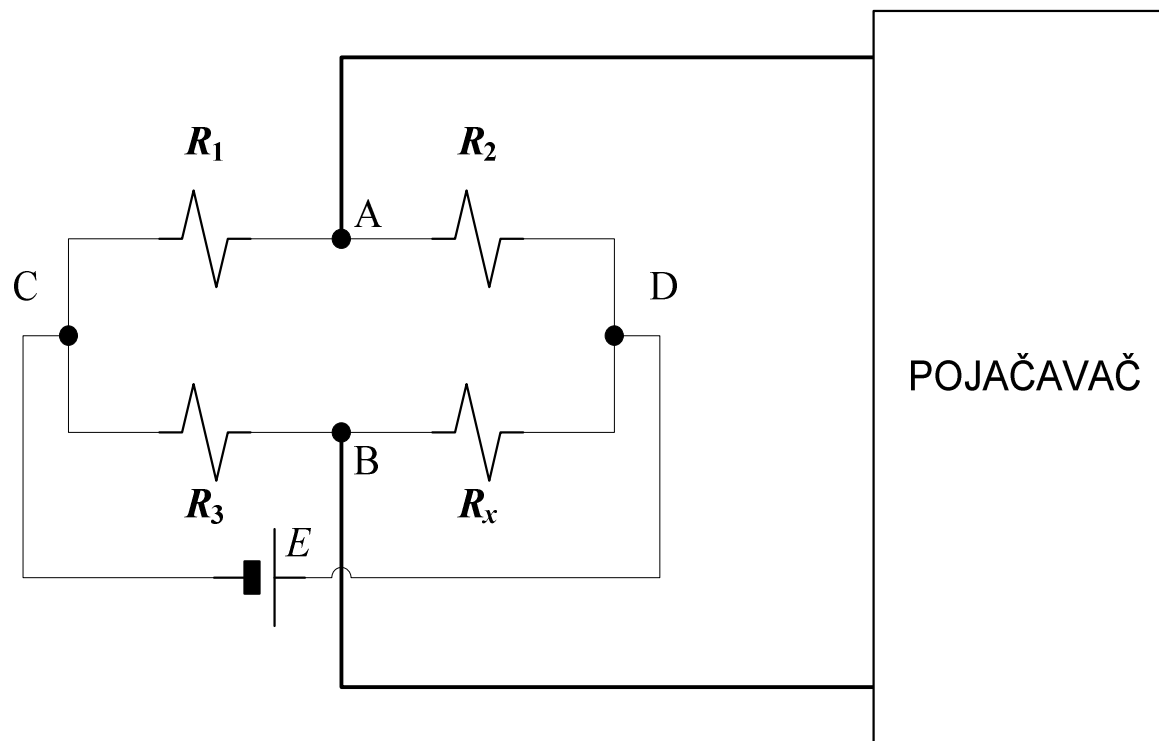
Senzor
(pretvarač neelektrične veličine
u otpornost)

Lab. vežba: upotreba NTC termistora kao senzora temperature



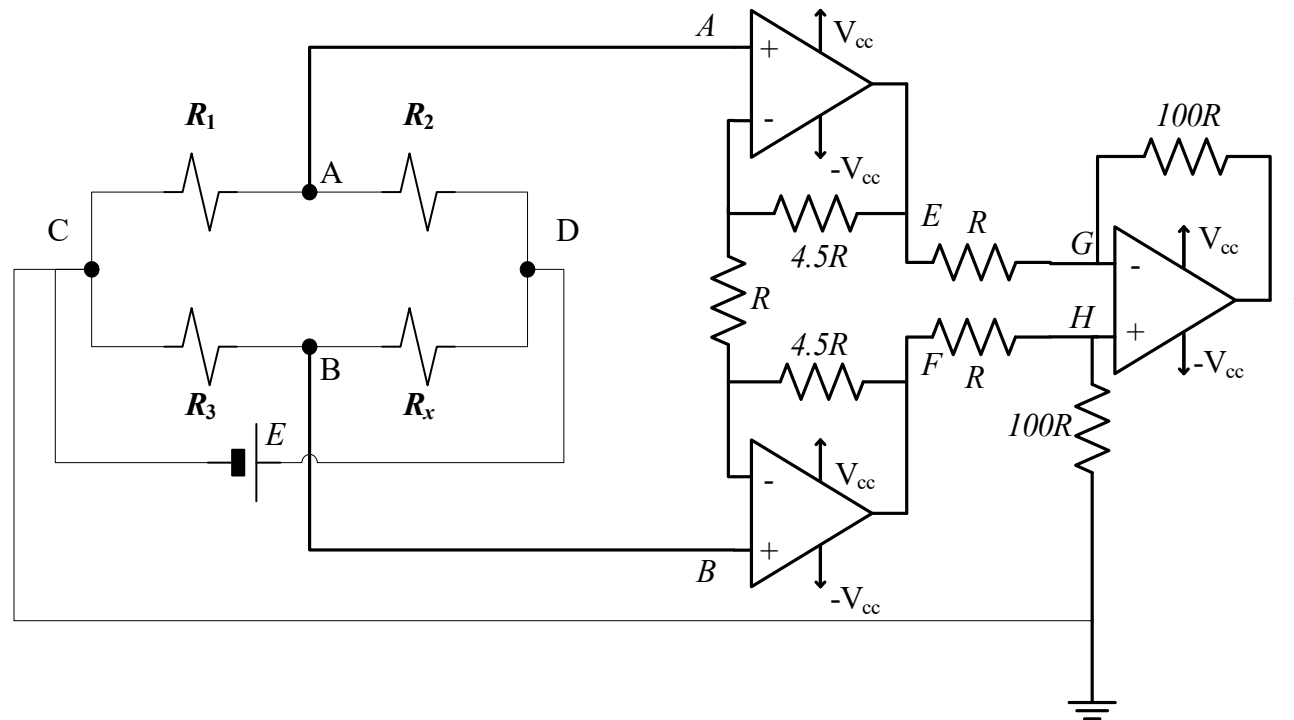
MERENJE OTPORNOSTI MOSTOM

Vitstonov most – **neuravnoteženi** - PRIMENA



u_{AB} je obično neophodno pojačati pre merenja!

Primer

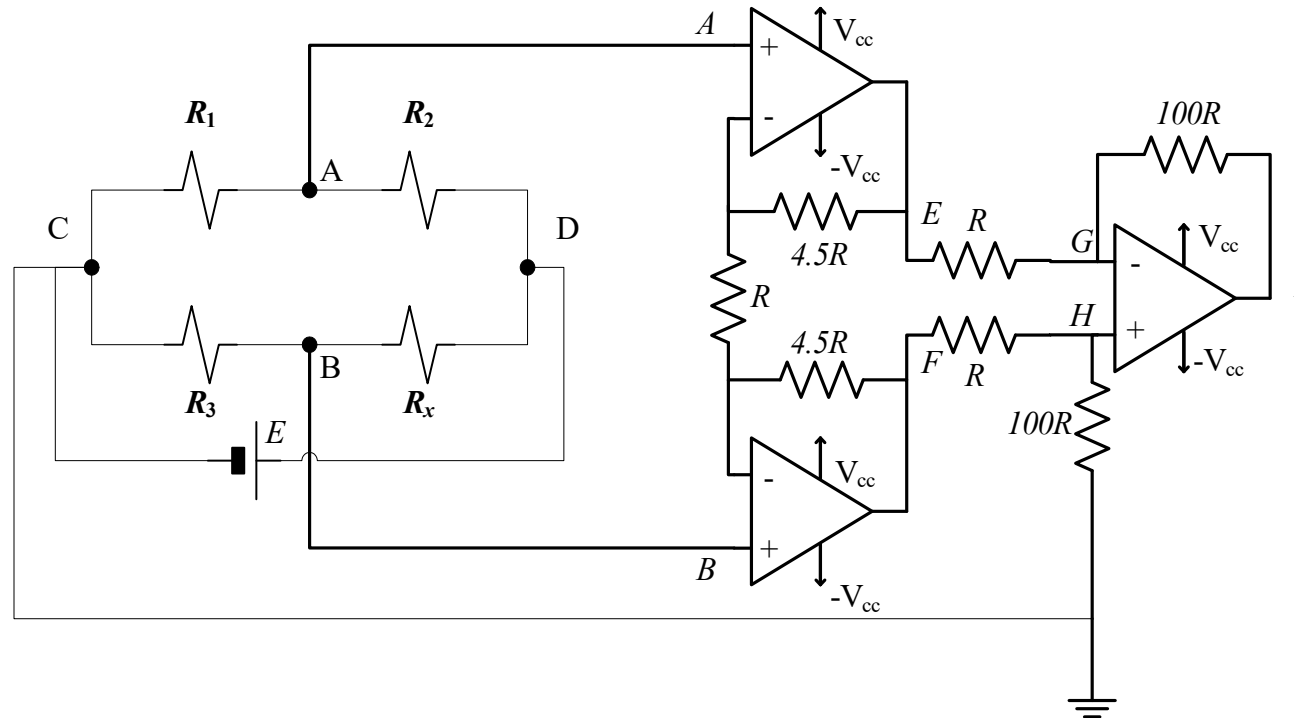


Na slici je prikazan **neuravnotežen** Vitstonov most čiji izlazni napon u_{AB} se vodi na instrumentacioni pojačavač. Smatrati da je indikator idealan voltmetar. R_x je promenljiva otpornost. Poznato je:

$$R_1 = R_2 = R_3 = R$$
$$R_x = R + \Delta R_x, \Delta R_x / R \ll 1$$

Odrediti vrednost izlaznog napona pojačavača v_i u funkciji elemenata kola.

Primer



Na slici je prikazan **neuravnotežen** Vitstonov most čiji izlazni napon u_{AB} se vodi na instrumentacioni pojačavač. Smatrati da je indikator idealan voltmetar. R_x je promenljiva otpornost. Poznato je:

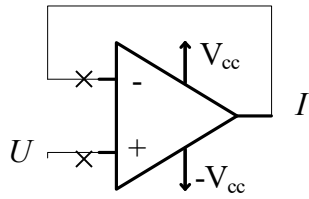
$$R_1 = R_2 = R_3 = R$$
$$R_x = R + \Delta R_x, \Delta R_x / R \ll 1$$

Odrediti vrednost izlaznog napona pojačavača v_i u funkciji od R , E i ΔR_x .

Na prethodnim stranicama je izveden izraz: $u_{AB} \approx E \cdot \frac{\Delta R_x}{4R}$

Primer

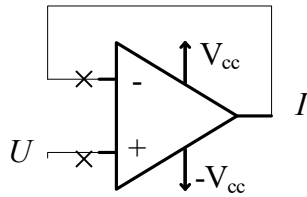
Digresija:



$$\Rightarrow v_1 = ?$$

Primer

Digresija:



Linearni režim rada operacionog pojačavača:

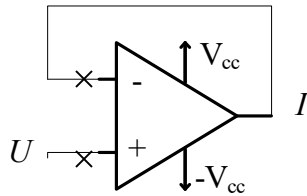
• ulazna struja je nula

• $v^+ = v^-$

$$v_I = \left\{ \begin{array}{l} v_U, \quad -V_{cc} < v_U < V_{cc} \\ V_{cc}, \quad v_U \geq V_{cc} \\ -V_{cc}, \quad v_U \leq -V_{cc} \end{array} \right\}$$

Primer

Digresija:



Linearni režim rada operacionog pojačavača:

- ulazna struja je nula
- $v^+ = v^-$

$$v_I = \left\{ \begin{array}{ll} v_U, & -V_{cc} < v_U < V_{cc} \\ V_{cc}, & v_U \geq V_{cc} \\ -V_{cc}, & v_U \leq -V_{cc} \end{array} \right\}$$

$$\frac{v_I - v_G}{100R} = \frac{v_G - v_E}{R}$$

$$v_I = 101v_G - 100v_E = 101v_H - 100v_E$$

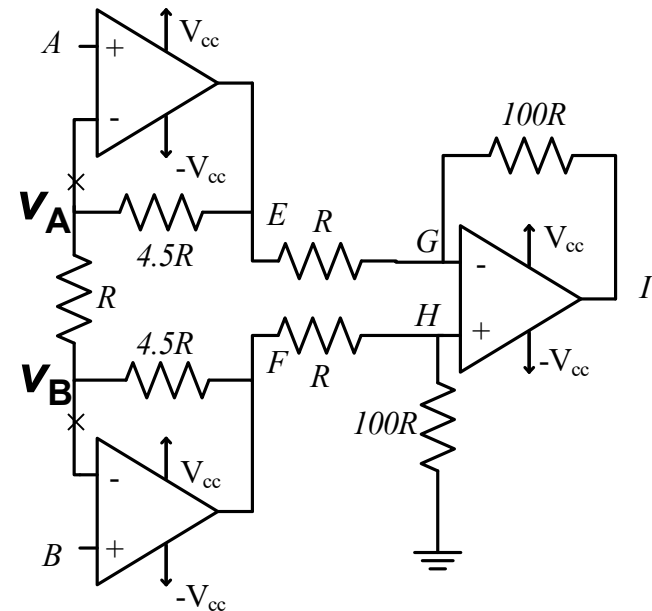
$$v_I = 101v_F \frac{100R}{R+100R} - 100v_E$$

$$v_I = 100(v_F - v_E)$$

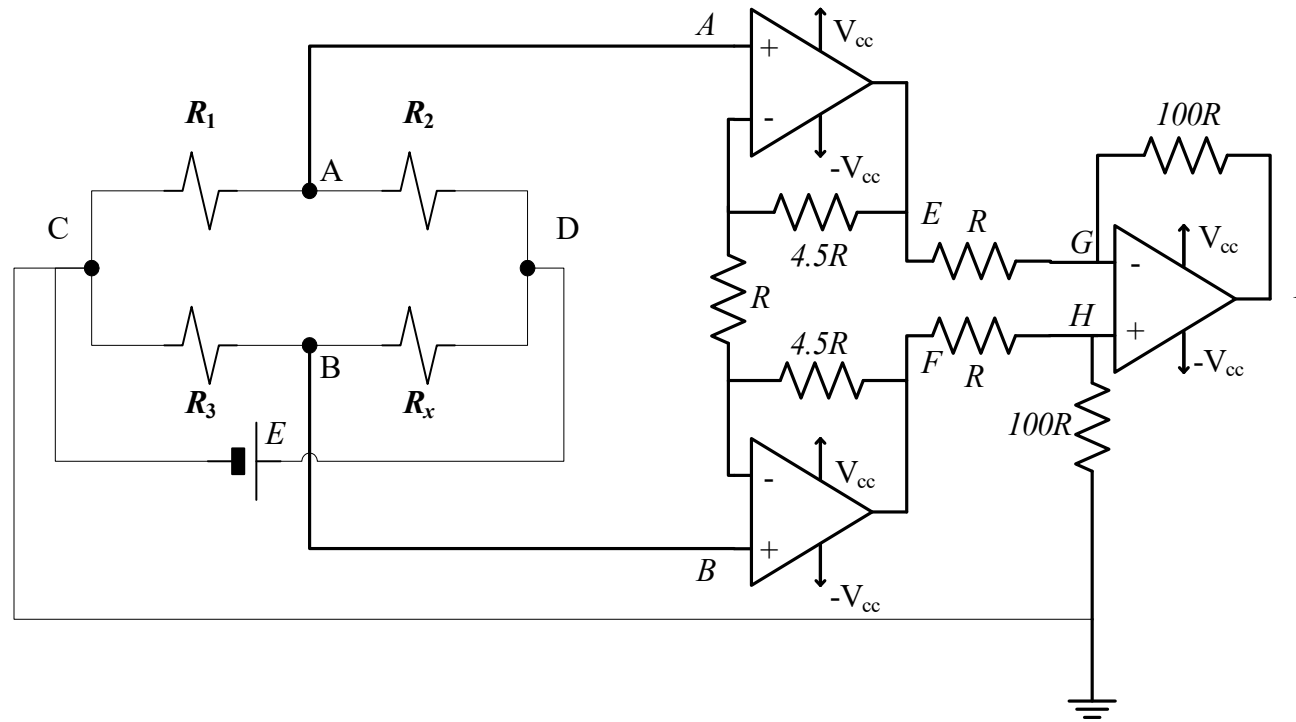
$$\frac{v_E - v_F}{10R} = \frac{v_A - v_B}{R}$$

$$v_E - v_F = 10(v_A - v_B)$$

$$v_I = -100 \cdot 10(v_A - v_B) = -1000 \cdot u_{AB} \approx -1000E \cdot \frac{\Delta R_x}{4R}$$



Primer



$$v_I = \left\{ \begin{array}{l} -1000E \cdot \frac{\Delta R_x}{4R}, \quad -V_{cc} < -1000E \cdot \frac{\Delta R_x}{4R} < V_{cc} \\ V_{cc}, \quad -1000E \cdot \frac{\Delta R_x}{4R} \geq V_{cc} \\ -V_{cc}, \quad -1000E \cdot \frac{\Delta R_x}{4R} \leq -V_{cc} \end{array} \right\}$$

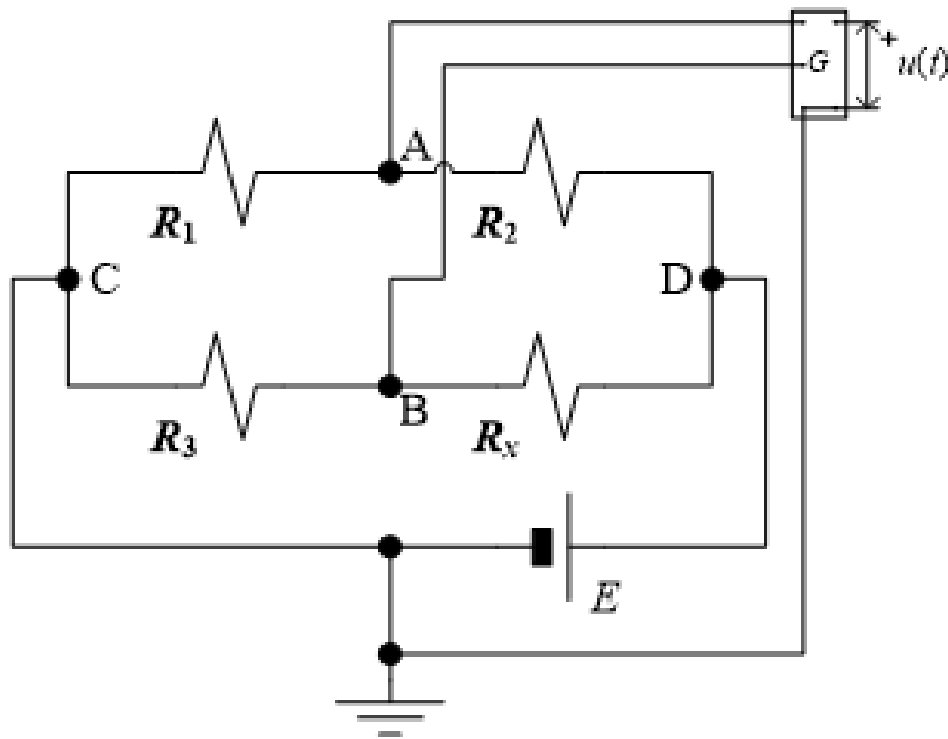
februar 2017

4. Na Sl. 4 je prikazan neuravnotežen Vitstonov most sa diferencijalnim pojačavačem pojačanja $G=100$ za merenje električne otpornosti R_x . Za merenje napona $u(t)=G \cdot u_{AB}(t)$ na izlazu pojačavača se koristi osciloskop.

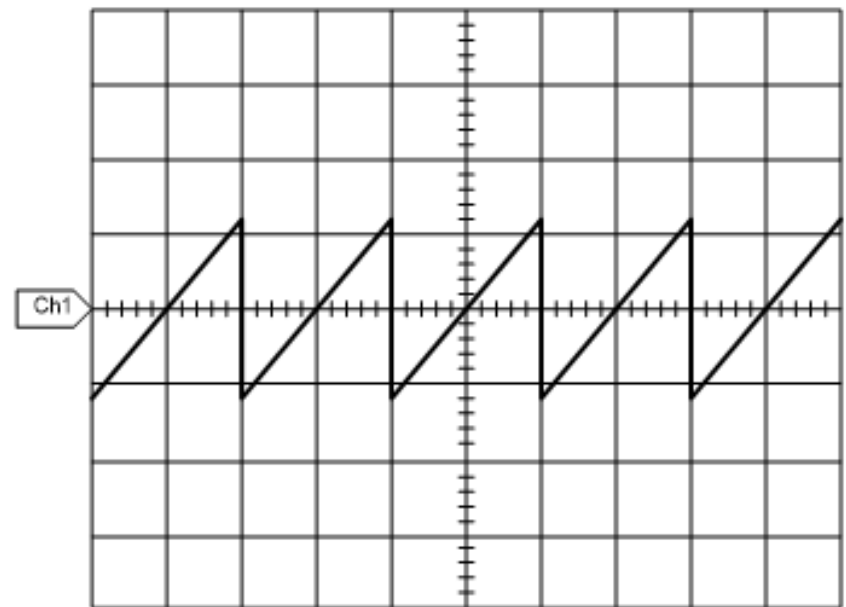
a) Odrediti izraz za napon u_{AB} u funkciji vrednosti otpornika R_1 , R_2 , R_3 i R_x .

b) Ako je R_x promenljivo i napon $u(t)$ ima oblik kao napon na kanalu 1 osciloskopa, Sl. 1, nacrtati grafik $R_x(t)$ za: $E=12\text{ V}$, $R_1=R_2=R_3=R=100\ \Omega$ i $R_x=R+\Delta R_x$ (uzeti u obzir da je $\Delta R_x/R \ll 1$).

c) Na koji način bi trebalo povezati osciloskop ako bi se direktno merio napon u_{AB} ?

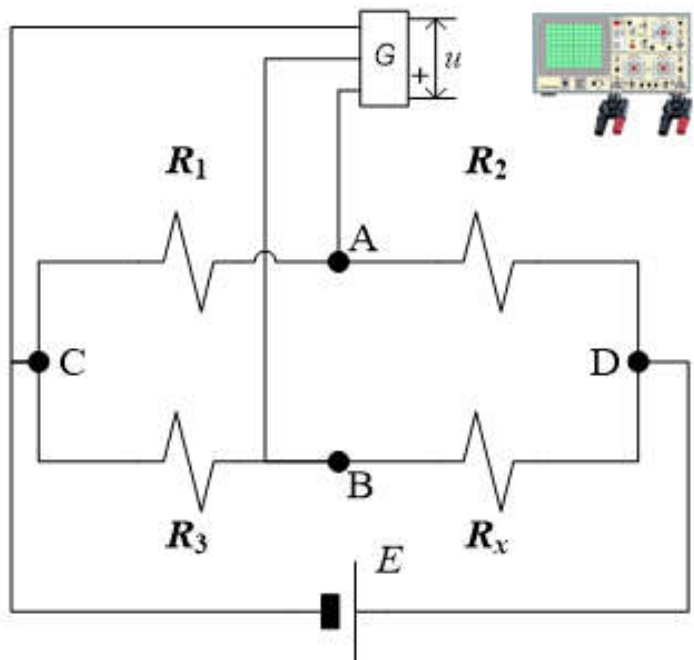


Sl. 4.

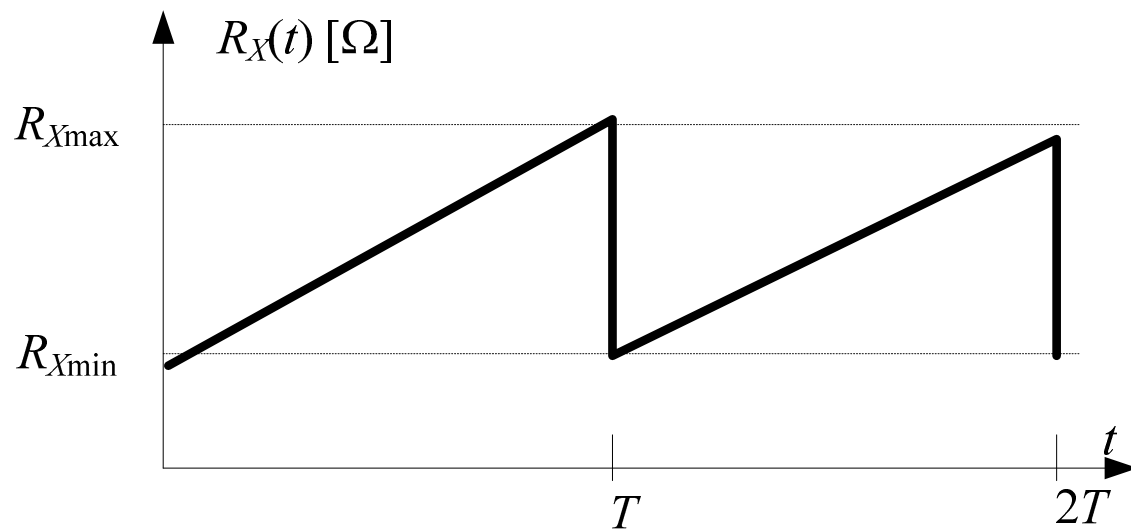


Sl. 1.

$k_y=2\text{V/div}$



Sl. 4.1



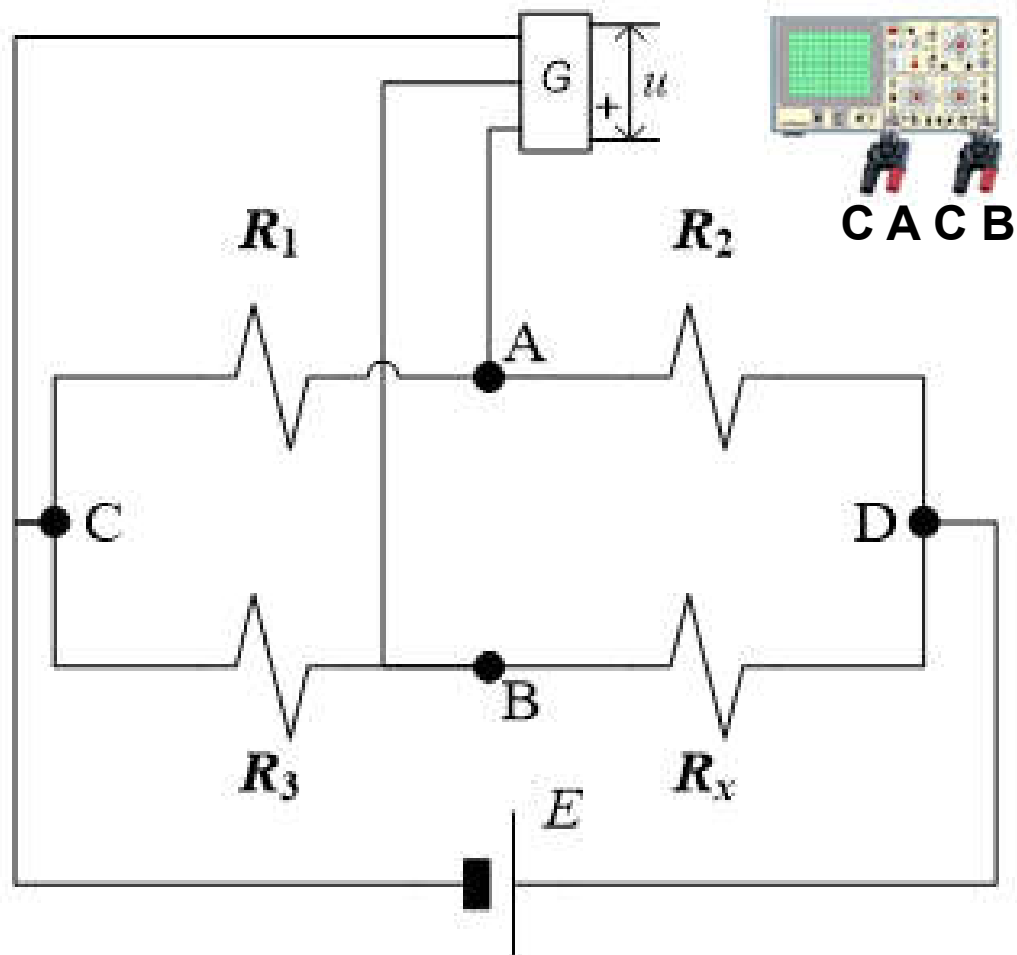
$$\text{a) } u_{AB} = u_{AD} + u_{DB} = -E \cdot \frac{R_2}{R_1 + R_2} + E \cdot \frac{R_x}{R_x + R_3} = E \cdot \frac{R_x R_1 - R_2 R_3}{(R_1 + R_2)(R_x + R_3)}$$

$$\text{b) } u_{AB} = E \cdot \frac{(R + \Delta R_x)R - R^2}{2R(R + \Delta R_x + R)} = E \cdot \frac{\Delta R_x}{2R \left(2 + \frac{\Delta R_x}{R}\right)} \approx E \cdot \frac{\Delta R_x}{4R}$$

$$R_x(t) = R + \Delta R_x(t) = R \cdot \left(1 + \frac{4u_{AB}(t)}{E}\right) = R \cdot \left(1 + \frac{4u(t)}{GE}\right)$$

$$R_{x\max} = 100.8 \, \Omega, \quad R_{x\min} = 99.2 \, \Omega$$

c)

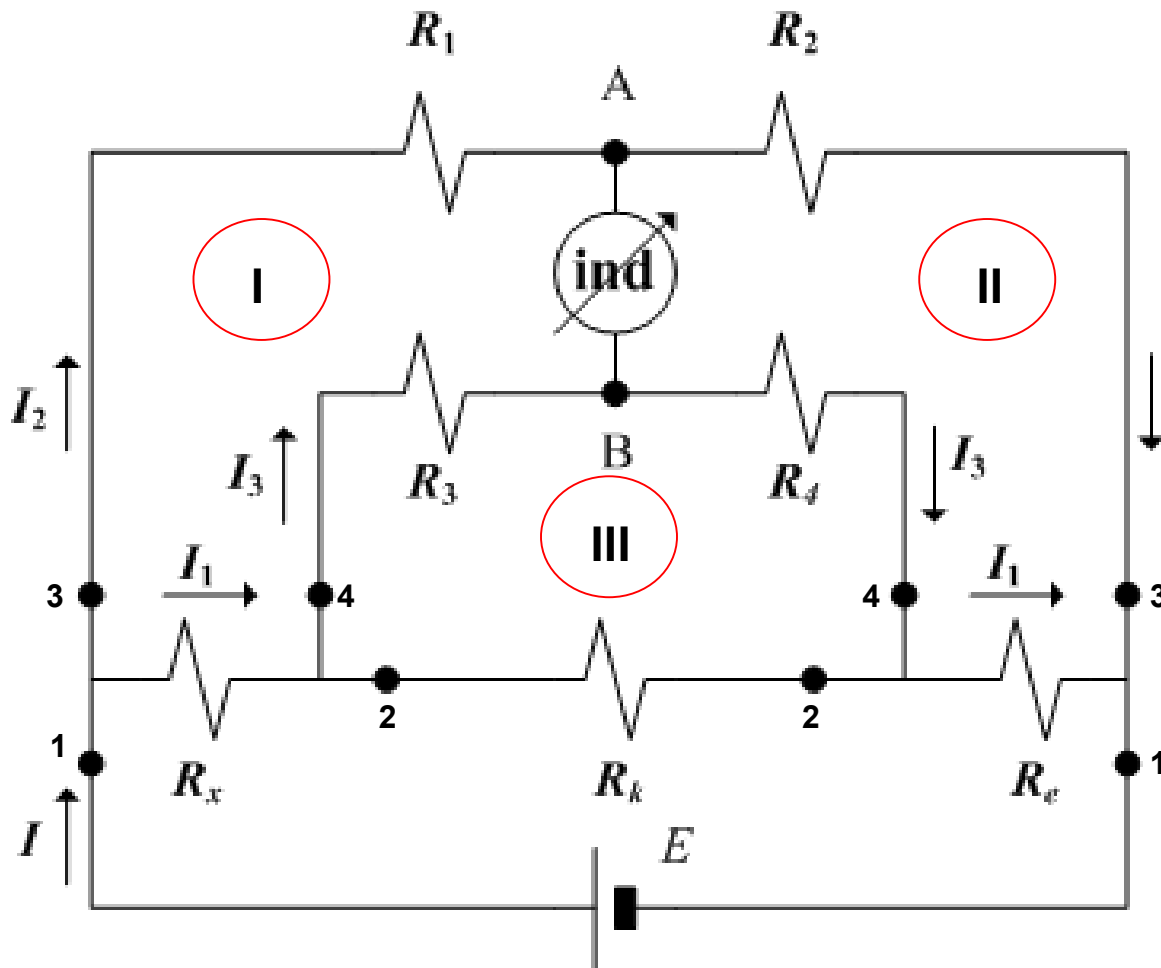


Sl. 4.

CH1+: tačka A, CH1-: tačka C, CH2+: tačka B, CH2-: tačka C
Invertovati kanal 2 (*Invert*) i izabrati opciju za sabiranje kanala (*ADD*).

MERENJE OTPORNOSTI MOSTOM

Kelvinov (ili Tomsonov) most



Kelvinov otpornik
(otpornik sa 4 pristupa)
1,2 – strujni pristupi
3,4 – naponski pristupi

$$I \quad I_2 R_1 = I_1 R_x + I_3 R_3$$

$$II \quad I_2 R_2 = I_1 R_e + I_3 R_4$$

$$III \quad I_3 (R_4 + R_3) = (I_1 - I_3) R_k$$

$$R_x = R_e \frac{R_1}{R_2} + \frac{R_k (R_1 R_4 - R_2 R_3)}{R_2 (R_3 + R_4 + R_k)}$$

Ako je: $R_1 R_4 = R_2 R_3$



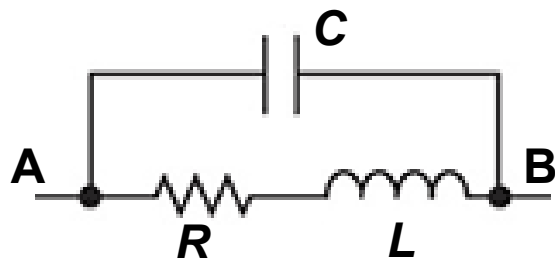
$$R_x = R_e \frac{R_1}{R_2}$$

R_k - otpornost mernih vodova i kontakata

R_e - etalonski otpornik

Merni opseg (male otpornosti): od $1 \mu\Omega$ do 1Ω .

EKVIVALENTAN MODEL OTPORNOSTI



Ekvivalentna šema realnog otpornika

$$\underline{Z}_{AB} = R_{ekv} + jX_{ekv}$$
$$\underline{Z}_{AB} = \frac{R + j\omega[L(1 - \omega^2 LC) - R^2 C]}{(1 - \omega^2 LC)^2 + (\omega RC)^2}$$

Niska učestanost

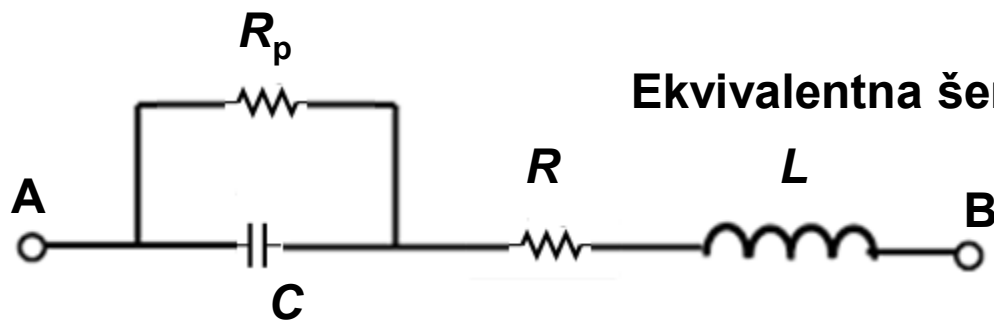
Kada je $1 - \omega^2 LC \approx 1$ i $1 + (\omega RC)^2 \approx 1$:

$$\underline{Z}_{AB} = R + j\omega[L - R^2 C]$$

Za malo R: $\underline{Z}_{AB} \approx R + j\omega L$

Za veliko R: $\underline{Z}_{AB} \approx R - j\omega R^2 C$

EKVIVALENTAN MODEL KAPACITIVNOSTI



Ekvivalentna šema realnog kondenzatora

R – otpornost metalnih delova kondenzatora

L – induktivnost provodnika i elektroda

R_p – dielektrični gubici

$$\underline{Z}_{AB} = R + j\omega L + \frac{R_p}{1 + j\omega C R_p}$$

$$\underline{Z}_{AB} = R + j\omega L + \frac{R_p - j\omega C R_p^2}{1 + (\omega C R_p)^2}$$

Važi: $1 + (\omega C R_p)^2 \approx (\omega C R_p)^2$

$$\underline{Z}_{AB} = R + j\omega L + \frac{R_p - j\omega C R_p^2}{(\omega C R_p)^2}$$

$$\underline{Z}_{AB} = R + \frac{1}{\omega^2 C^2 R_p} + j\left(\omega L - \frac{1}{\omega C}\right)$$

$$\underline{Z}_{AB} = R_{ekv} + jX_{ekv}$$

$$R_{ekv} = R + \frac{1}{\omega^2 C^2 R_p}, \quad X_{ekv} = \omega L - \frac{1}{\omega C}$$

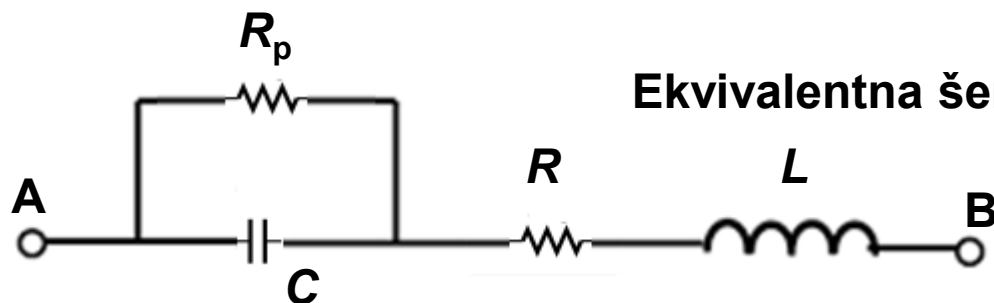
$X_{ekv} = 0$ za $\omega = \omega_0 = 1/\sqrt{LC}$ (rezonancija kola)

Za $\omega = \omega_0$ $\underline{Z}_{AB} = R_{ekv}$

Za $\omega > \omega_0$ induktivna komponenta je dominantna

Za $\omega < \omega_0$ kapacitivna komponenta je dominantna

EKVIVALENTAN MODEL KAPACITIVNOSTI



Ekvivalentna šema realnog kondenzatora

- R – otpornost metalnih delova kondenzatora
- L – induktivnost provodnika i elektroda
- R_p – dielektrični gubici

$$\underline{Z}_{AB} = R_{ekv} + jX_{ekv}$$

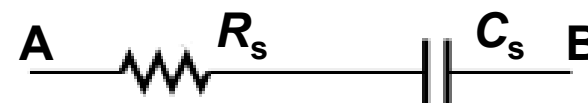
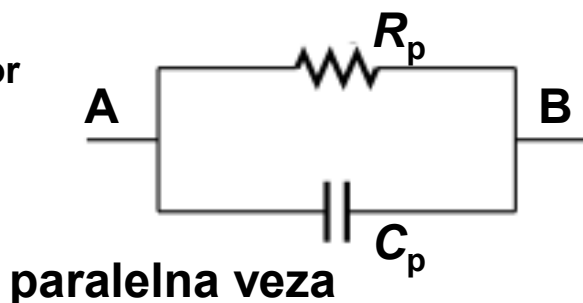
$$R_{ekv} = R + \frac{1}{\omega^2 C^2 R_p}, \quad X_{ekv} = \omega L - \frac{1}{\omega C}$$

Za $\omega < \omega_0$ kapacitivna komponenta je dominantna



uprošćeni modeli kapacitivnosti

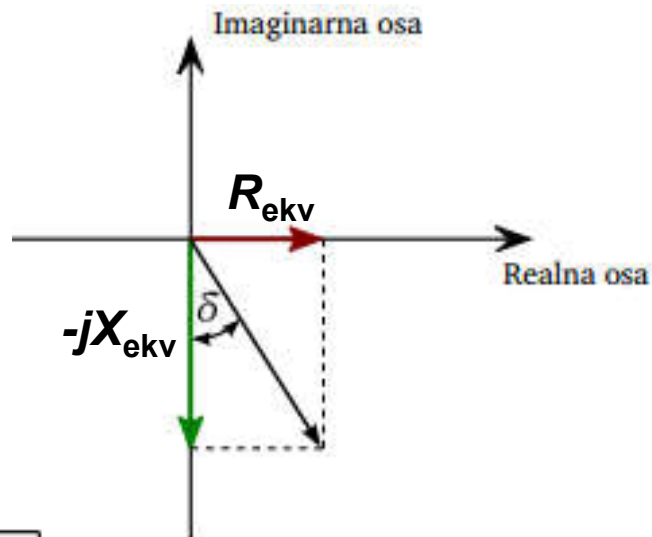
Dobar izbor u slučaju velike reaktanse



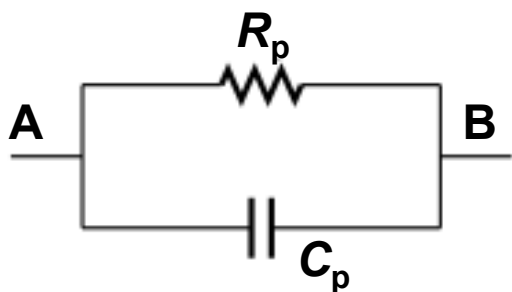
Dobar izbor u slučaju male reaktanse

EKVIVALENTAN MODEL KAPACITIVNOSTI

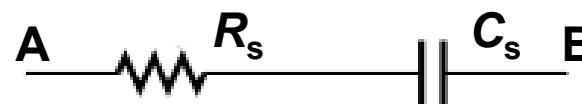
Faktor disipacije (tangens ugla gubitaka) se definiše kao: $\text{tg}\delta = \frac{R_{ekv}}{X_{ekv}}$



$$\underline{Z}_{AB} = R_{ekv} - jX_{ekv}$$



paralelna veza



redna veza

$$\underline{Z}_{AB} = \frac{R_p \frac{j}{\omega C_p}}{R_p + \frac{j}{\omega C_p}} = \frac{R_{ekv}}{1 + (\omega C_p R_p)^2} - j \frac{X_{ekv}}{1 + (\omega C_p R_p)^2}$$

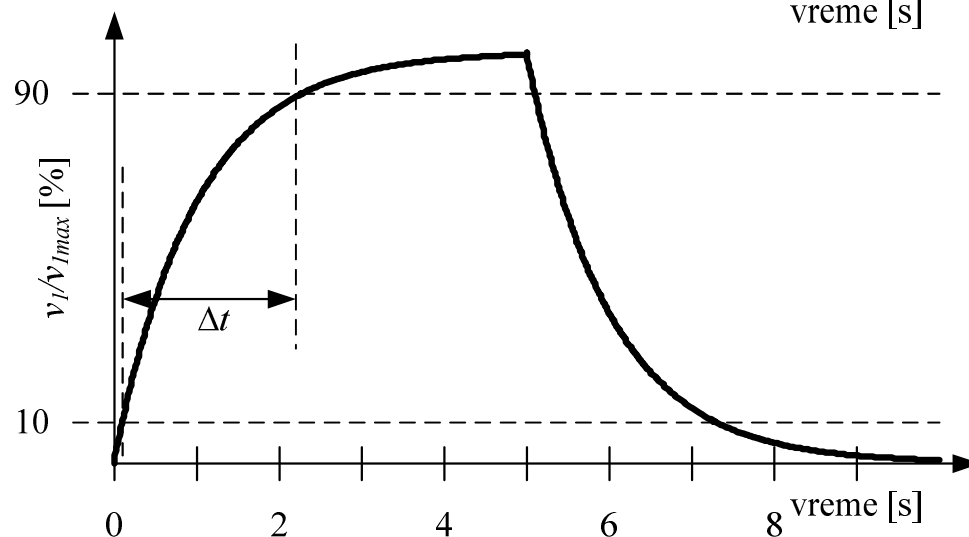
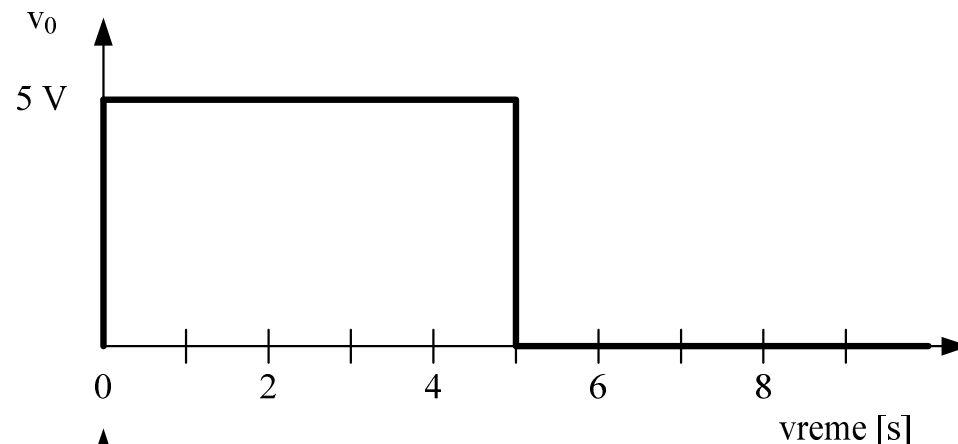
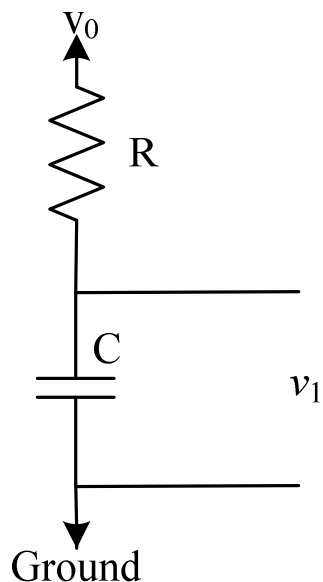
$$\text{tg}\delta = \frac{R_{ekv}}{X_{ekv}} = \frac{1}{\omega C_p R_p}$$

$$\underline{Z}_{AB} = R_s - j \frac{1}{\omega C_s}$$

$$R_{ekv} = R_s \quad X_{ekv} = \frac{1}{\omega C_s}$$

$$\text{tg}\delta = \frac{R_{ekv}}{X_{ekv}} = \omega C_s R_s$$

ODZIV KONDENZATORA NA IMPULSNU POBUDU (lab. vežba)



$$v_1 = v_0 \left(1 - e^{-\frac{t}{\tau}} \right),$$

$$0 < t < 5 \text{ s}$$

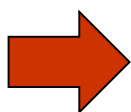
$$v_1 = v_0 e^{-\frac{t-T}{\tau}},$$

$$5 \text{ s} \leq t < 10 \text{ s}, T = 5 \text{ s}$$

$$v_1 = 0,$$

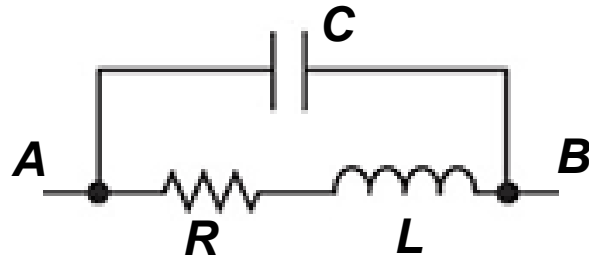
$$t > 10 \text{ s}$$

$$C = \frac{\Delta t}{R \ln 9} = \frac{t_{90} - t_{10}}{R \ln 9}$$



Procena kapacitivnosti na osnovu merenja trajanja uzlazne ivice

EKVIVALENTAN MODEL KALEMA



Ekvivalentna šema realnog kalema

$$\underline{Z}_{AB} = R_{ekv} + jX_{ekv}$$
$$\underline{Z}_{AB} = \frac{R + j\omega[L(1 - \omega^2 LC) - R^2 C]}{(1 - \omega^2 LC)^2 + (\omega RC)^2}$$

Kod kalemova se nastoji da R i C budu što manji, pa je:

Kada je $R^2 C \ll L(1 - \omega^2 LC)$ i $(\omega RC)^2 \ll (1 - \omega^2 LC)^2$:

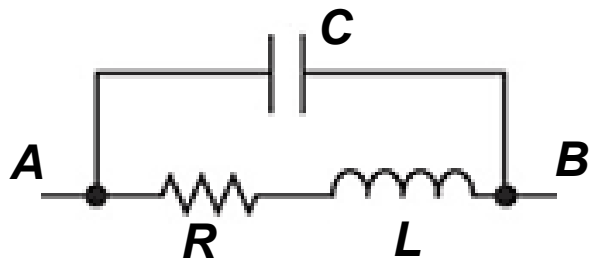
$$\underline{Z}_{AB} = \frac{R}{(1 - \omega^2 LC)^2} + j\omega \frac{L}{1 - \omega^2 LC} = R_{ekv} + jX_{ekv}$$

$$R_{ekv} = \frac{R}{(1 - \omega^2 LC)^2} \quad X_{ekv} = \omega \frac{L}{1 - \omega^2 LC} = \omega L_{ekv}, \quad L_{ekv} = \frac{L}{1 - \omega^2 LC}$$

$f \uparrow \Rightarrow R_{ekv} \uparrow, L_{ekv} \uparrow$ sve do $f = f_0$

$f = f_0 = 1/2\pi\sqrt{LC}$ rezonantna frekvencija

EKVIVALENTAN MODEL KALEMA

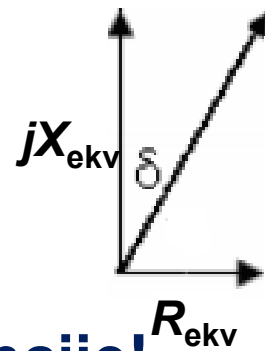


Ekvivalentna šema realnog kalema

$$R_{ekv} = \frac{R}{(1 - \omega^2 LC)^2} \quad X_{ekv} = \omega \frac{L}{1 - \omega^2 LC}$$

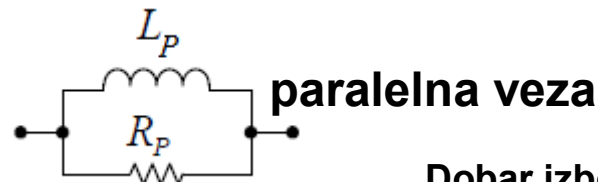
Faktor dobrote realnog kalema Q

(odnos maksimalne akumulisane energije prema disipiranoj energiji):



$$Q = \frac{1}{\operatorname{tg} \delta} = \frac{X_{ekv}}{R_{ekv}} = \frac{\omega L}{R} (1 - \omega^2 LC) \quad \text{Nelinearna funkcija frekvencije!}$$

↓ uprošćene šeme induktivnosti



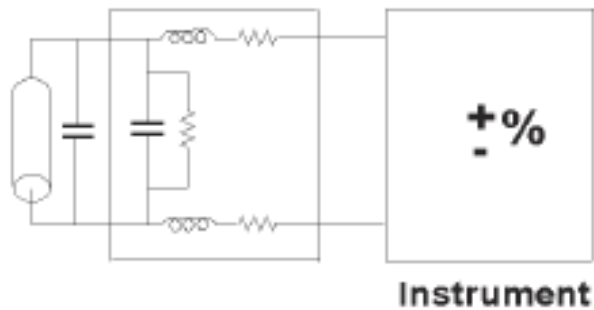
Dobar izbor u slučaju male reaktanse

$$Q = \frac{X_{ekv}}{R_{ekv}} = \frac{\omega L_s}{R_s}$$

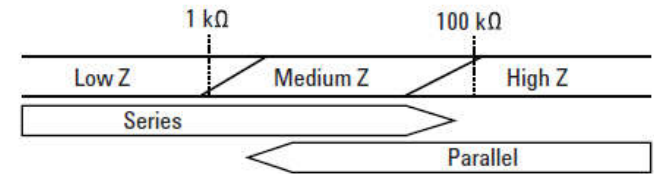
$$Q = \frac{X_{ekv}}{R_{ekv}} = \frac{R_p}{\omega L_p}$$

Dobar izbor u slučaju velike reaktanse

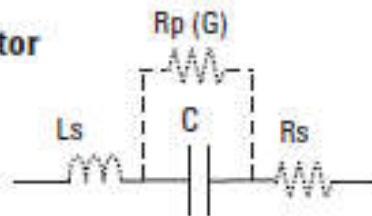
Merenje impedanse LRC metrom



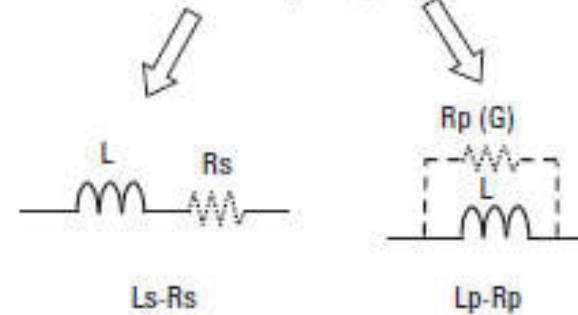
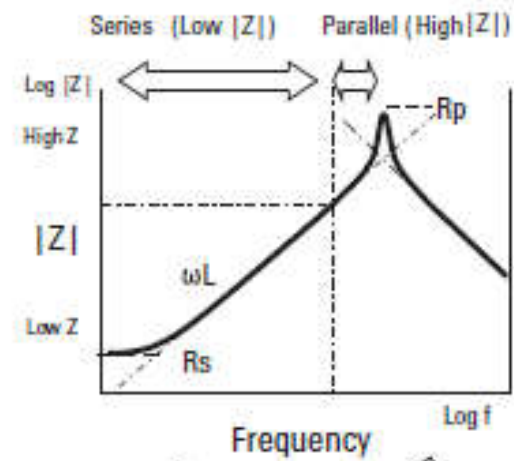
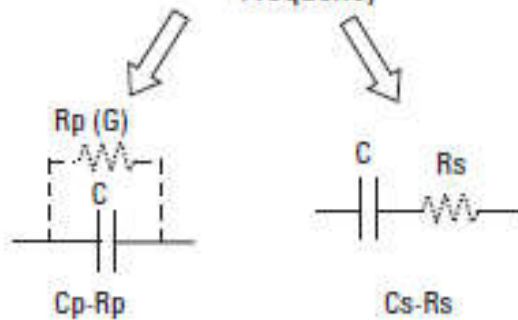
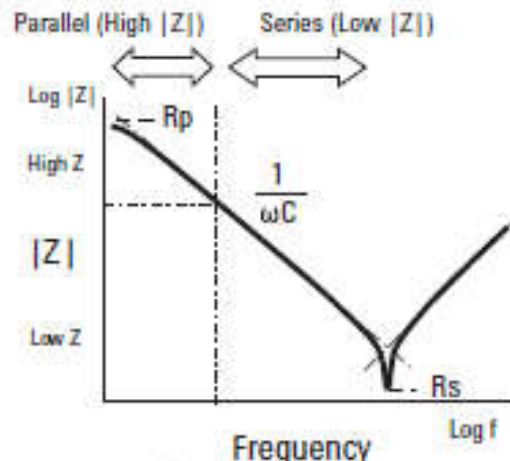
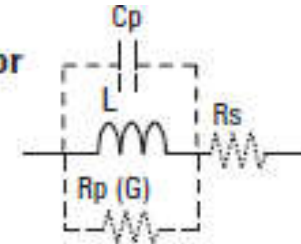
LRC meter



(a) Capacitor



(b) Inductor

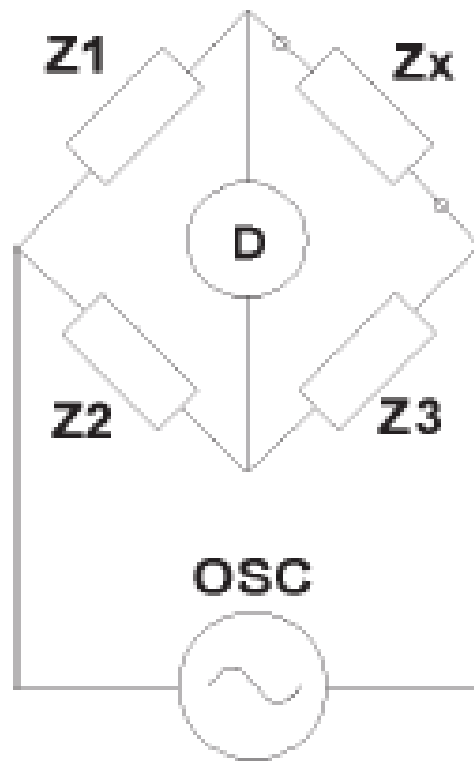


LRC metar – princip rada

- Merni most
- Rezonantni metod
- Merenje struje i napona
- Automatizovani most (autobalancing bridge)

LRC metar

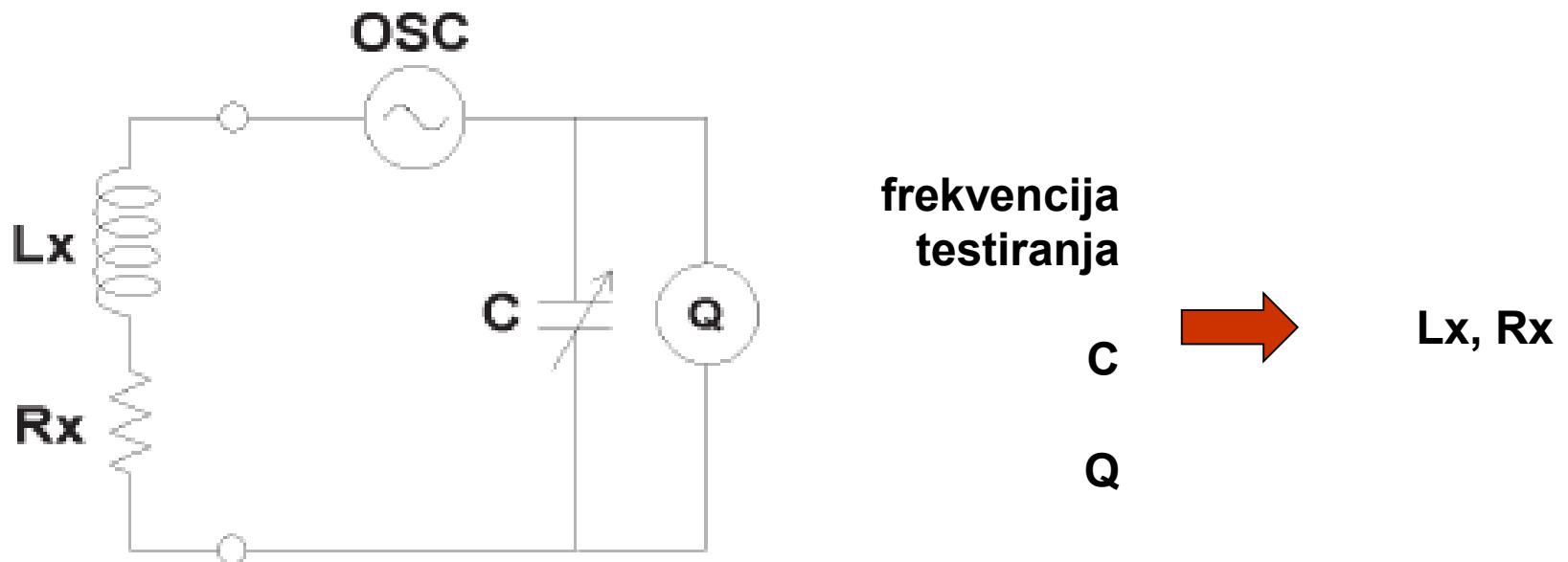
Most za naizmeničnu struju



$$Z_x = \frac{Z_1}{Z_2} Z_3$$

LRC metar

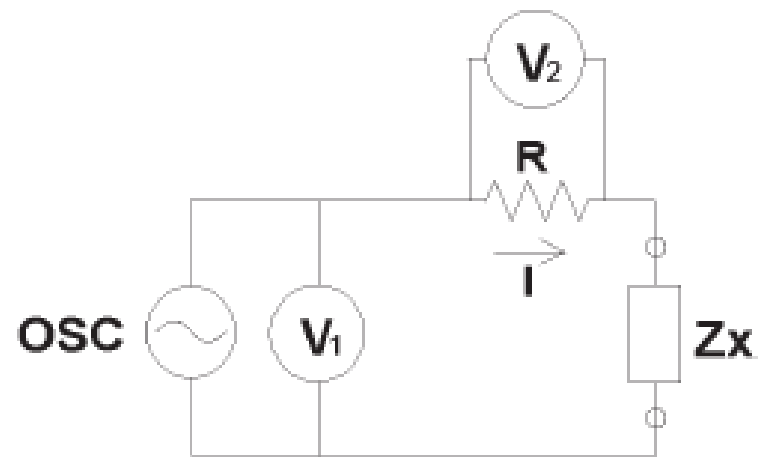
Rezonantni metod



Podešavanjem C se kolo dovede u rezonanciju.

LRC metar

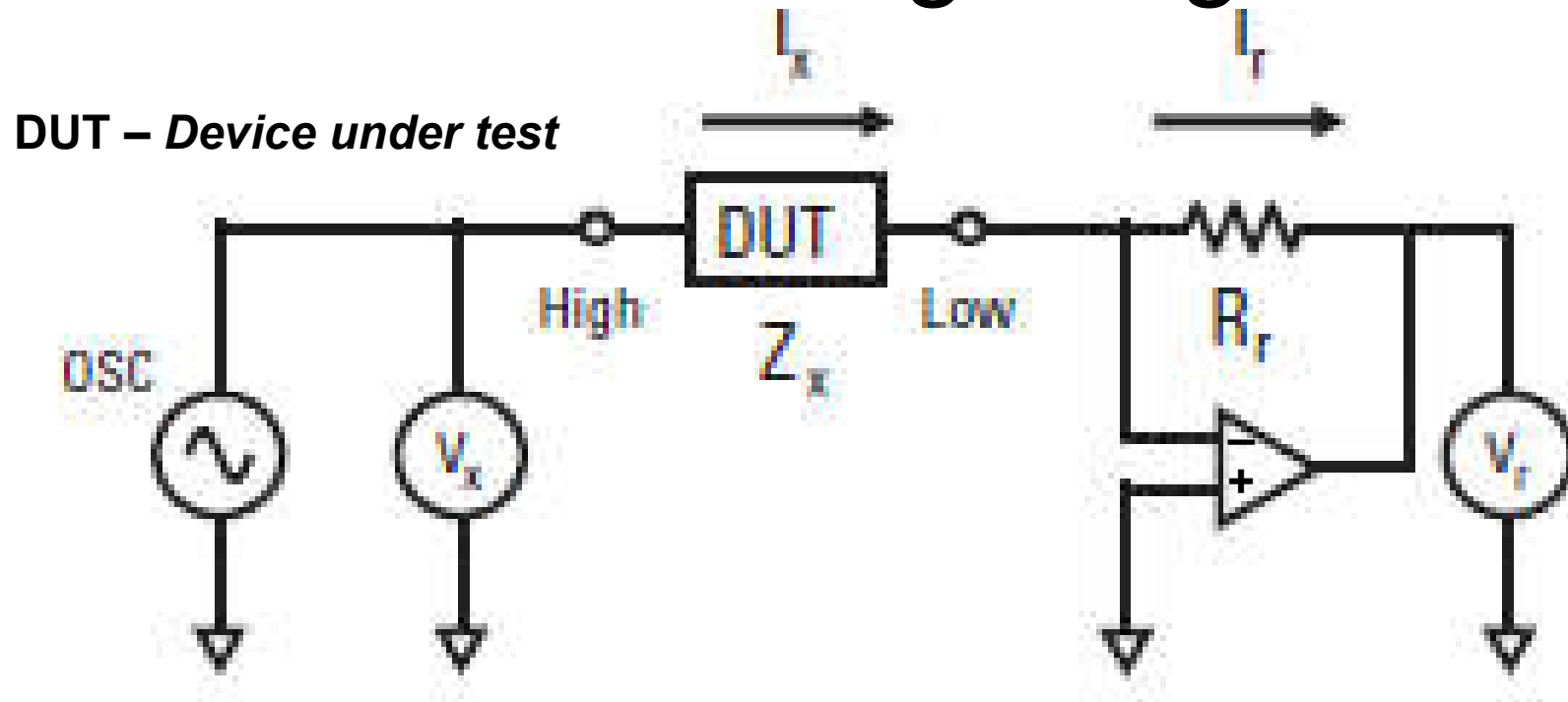
Merenje struje i napona



$$Z_x = \frac{V_1}{I} = \frac{V_1}{V_2} R$$

LRC meter

Autobalancing bridge



$$\frac{V_x}{Z_x} = I_x = I_r = \frac{V_r}{R_r}$$

$$\rightarrow Z_x = \frac{V_x}{I_x} = R_r \frac{V_x}{V_r}$$

LRC meter

	Advantages	Disadvantages	Applicable frequency range	Agilent measurement instruments	Common applications
Bridge method	<ul style="list-style-type: none"> • High accuracy (0.1% typ.) • Wide frequency coverage by using different types of bridges • Low cost 	<ul style="list-style-type: none"> • Needs to be manually balanced • Narrow frequency coverage with a single instrument 	DC to 300 MHz	None	Standard lab
Resonant method	<ul style="list-style-type: none"> • Good Q accuracy up to high Q 	<ul style="list-style-type: none"> • Needs to be tuned to resonance • Low impedance measurement accuracy 	10 kHz to 70 MHz	None	High Q device measurement
I-V method	<ul style="list-style-type: none"> • Grounded device measurement • Suitable to probe-type test needs 	<ul style="list-style-type: none"> • Operating frequency range is limited by transformer used in probe 	10 kHz to 100 MHz	None	Grounded device measurement
Auto-balancing bridge method	<ul style="list-style-type: none"> • Wide frequency coverage from LF to HF • High accuracy over a wide impedance measurement range • Grounded device measurement 	<ul style="list-style-type: none"> • Higher frequency ranges not available 	20 Hz to 110 MHz	E4980A E4981A 4294A 4294A+42941A ¹ 4294A+42942A ¹	Generic component measurement 1. Grounded device measurement

Agilent Impedance Measurement Handbook, A guide to measurement technology and techniques